

Innovations in EDA Webcast Series

How to Design an X-band MMIC PA

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 **Microwave
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Stuart Glynn
Power Amplifier Design Engineer



Plextek RFI



Plextek RFI

How to Design an X-band MMIC PA

Stuart Glynn and Liam Devlin



Plextek RFI

Introduction

- Target specification and application
- Design approach
- Device size and bias selection
- Load-pull simulation to select terminating impedances
- Circuit schematic
- Layout
- EM simulation and layout optimisation
- Final simulated performance
- Fabrication
- Measured performance
- Summary/Conclusions

Target Specification

Intended application: phased array radar

Parameter	Units	Target
Frequency	GHz	9 to 11
SS Gain	dB	> 12
P_{sat}	W	> 6
	dBm	> 37.7
PAE at P_{sat}	%	> 40
IRL	dB	> 9
ORL	dB	> 8
Drain Voltage	V	25

Design Approach

1. Device Level Simulations, (including load pull) to Identify:
 - Unit Device Size
 - Bias Point
 - Fundamental and Harmonic Impedance Targets (Load / Source)
 - Number of Stages
 - Number of Devices in Output Stage
2. Schematic Design
 - Matching Network Topology
 - Bias Network Topology
3. Layout
 - Preliminary Layout
 - EM Simulation and Layout Optimization
 - DRC

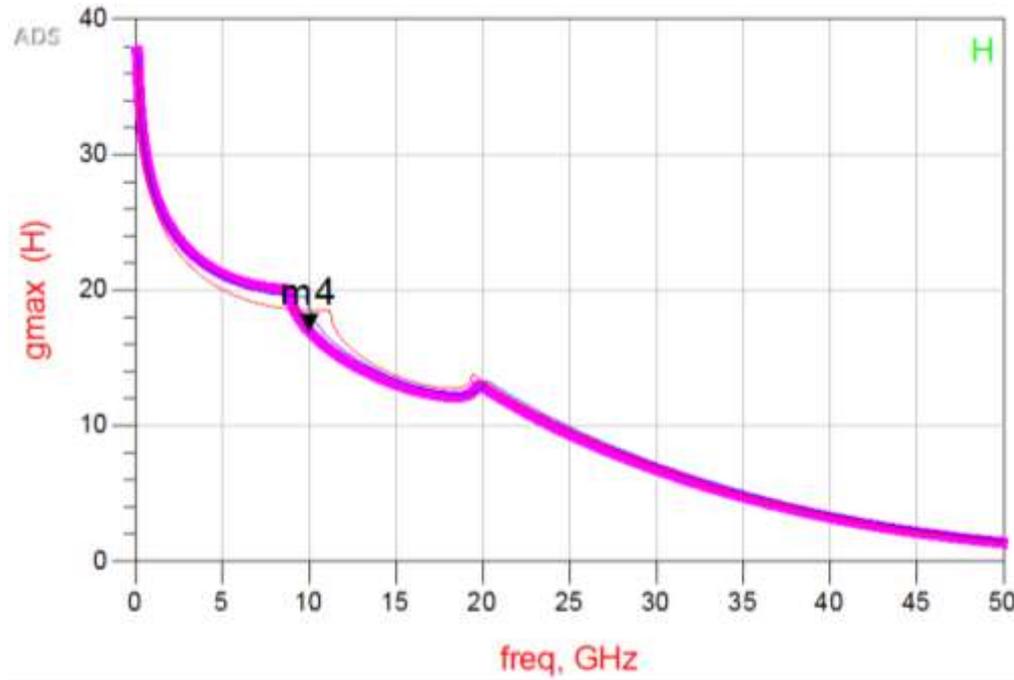
Device Selection

Gmax of three, 8 finger devices, biased at 100mA/mm from 25V. Each has a different finger width.

Red is for an 8x75 μ m device,

Blue is for an 8x125 μ m device,

Magenta is for an 8x150 μ m device



m4
freq=10.00GHz
gmax=16.925

- Largest device provides most RF power but must have adequate available gain
- Performance trade-off compared (gain, power and impedances) for different transistors
- The 8x150 μ m device was selected - also desirable as the break points in the Gmax plot indicate that it is unconditionally stable from 9GHz to 20GHz

Device Bias Point Selection

Vdd (V)	Ids/mm (mA)	P _{4dB} (dBm)	PAE at P _{4dB} (%)	Gain at P _{4dB} (dB)
25	100	36.44	52.47	11.44
25	75	36.56	53.01	11.06
25	50	36.67	53.45	10.67

Load and Source Pull analysis was carried out on the selected 8x150 μ m device at various current densities for V_{ds}=25V. Key parameters such as power, gain and power added efficiency (PAE) at 4dB compression were assessed.

Quiescent bias points of 75mA/mm and 37.5mA/mm were considered during the design, corresponding to a transistor bias of 90mA and 45mA at 25V.

Device Level Simulations

One Tone Load Pull Simulation with input power sweep; output power and PAE found at each fundamental or harmonic load

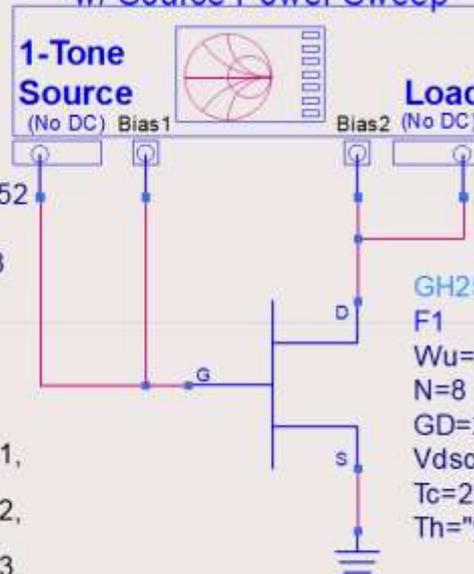
Load Pull Instr PSweep ZSweep1
X2

V_Bias1=-3.165 V
V_Bias2=25 V
RF_Freq=10 GHz
Pavs_dBm_Start1=0
Pavs_dBm_Stop1=20
Pavs_dBm_Step1=2.5
Pavs_dBm_Start2=21
Pavs_dBm_Stop2=28
Pavs_dBm_Step2=0.25
Swept_Harmonic_Num=1
Z_Load_Baseband=50+j*0
Z_Load_Fund=10+j*0
Z_Load_2nd=1.41+j*107.2
Z_Load_3rd=0.29-j*20.71
Z_imag_min=5

Z_imag_max=55
Z_imag_num_pts=25
Z_real_min=5
Z_real_max=55
Z_real_num_pts=25
Z_Source_Fund=2.22+j*5.52
Z_Source_2nd=1000+j*0
Z_Source_3rd=4.01+j*41.8

If Swept_Harmonic_Num=1,
Z_Load_Fund is ignored.
If Swept_Harmonic_Num=2,
Z_Load_2nd is ignored.
If Swept_Harmonic_Num=3,
Z_Load_3rd is ignored.

Load Pull Instr ZSweep
w/ Source Power Sweep



Push into instrument subcircuit to see or modify bias network, if necessary.

GH25NHF_10
F1
Wu=150 um
N=8
GD=2.7
Vdsc=25
Tc=25
Th="yes"

UMS GH25
TechInclude

TechIncludeGH25
TechIncludeGH25



Load Pull on Selected Device at 10GHz at 4dB Compression

Power and PAE contours are shown at 10GHz at 4dB compression.

The corresponding load harmonic terminations are:-

$$1.41\Omega + j*107.2\Omega \text{ at } 2^{\text{nd}} \text{ Harmonic}$$

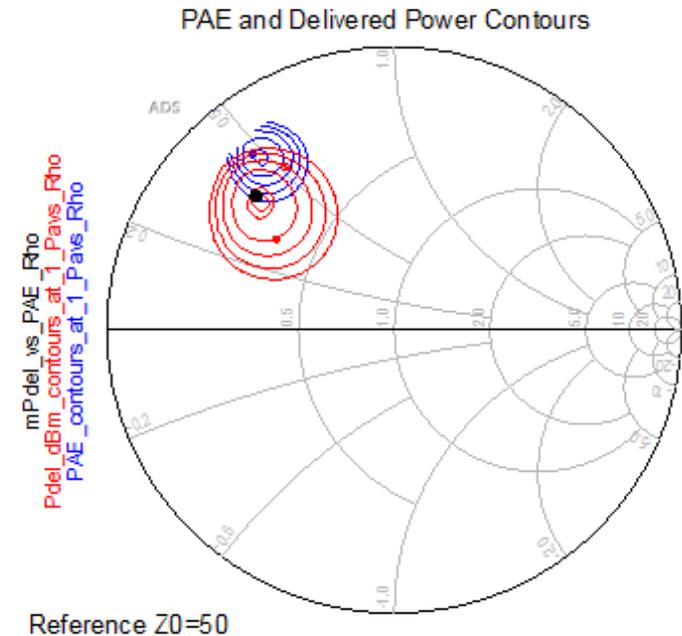
$$0.29\Omega - j*20.71\Omega \text{ at } 3^{\text{rd}} \text{ Harmonic}$$

The corresponding source impedances are:-

$$2.22\Omega + j*5.52\Omega \text{ at the fundamental}$$

$$1000\Omega + j0\Omega \text{ at } 2^{\text{nd}} \text{ Harmonic}$$

$$4.01\Omega + j*41.80\Omega \text{ at } 3^{\text{rd}} \text{ Harmonic}$$



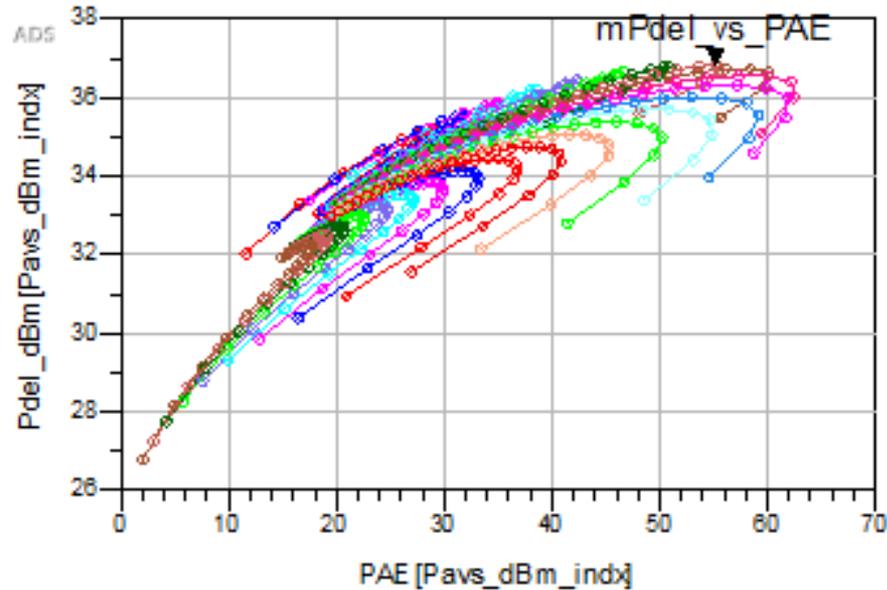
Maximum PAE
and contour
levels, %:

62.382
62.000
60.000
58.000
56.000
54.000

Maximum Power
Delivered and
contour levels, dBm:

36.785
36.750
36.500
36.250
36.000
35.750

Load Pull on Selected Device at 10GHz at 4dB Compression



- Above plot was used to select the fundamental load impedance that would be the best compromise between output power and PAE at 4dB compression
- The selected fundamental load Z target was $11.25\Omega + j*21.67\Omega$

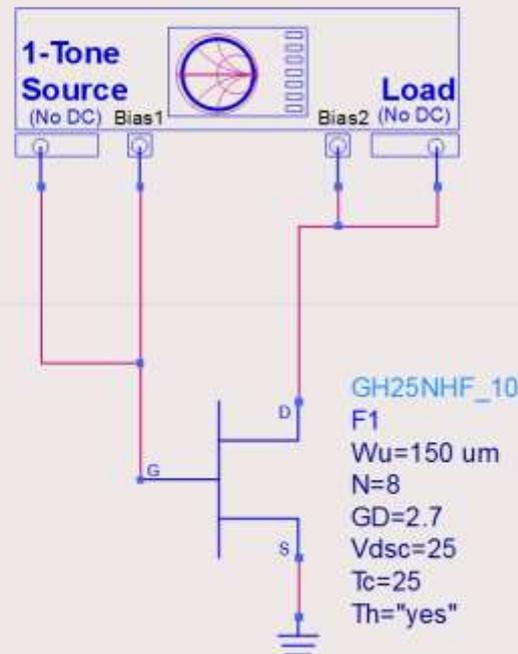
Device Level Simulations

One Tone Harmonic Load Phase Sweep;
output power and PAE found at each
fundamental or harmonic load

Harmonic_Load_Phase_Sweep_Instrument

X6
V_Bias1=-3.165 V
V_Bias2=25 V
RF_Freq=10 GHz
Pavs_dBm=25
Rho_Phase_Degrees_Start=0
Rho_Phase_Degrees_Stop=355
Rho_Phase_Degrees_Step=5
Z0=50+j*0
Specify_Load_S=no
Swept_Harmonic_Num=2
Z_Load_Baseband=50+j*0
Z_Load_Fund=11.25+j*21.67
Z_Load_2nd=1.41+j*107.2
Z_Load_3rd=0.29-j*20.71
Z_Load_4th=0+j*0
Z_Load_5th=0+j*0
Rho_Mag=0.99
Z_Source_Fund=2.22+j*5.52
Z_Source_2nd=1000+j*0
Z_Source_3rd=4.01+j*41.80

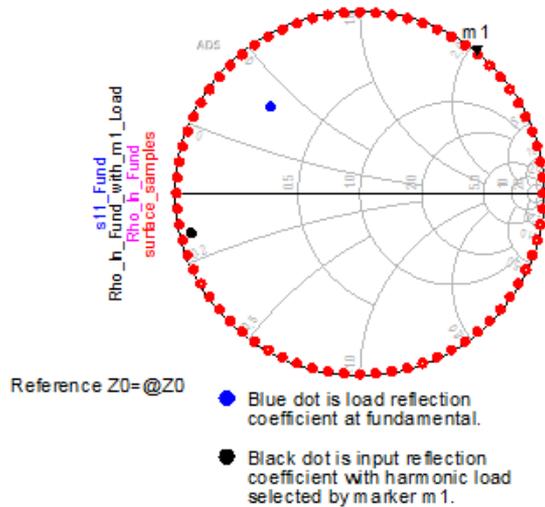
Load Phase Sweep Instrument



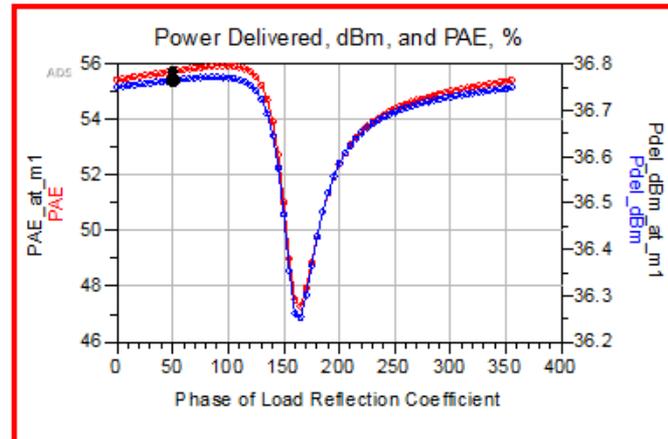
UMS GH25
TechInclude

TechIncludeGH25
TechIncludeGH25

Device Level Simulations – Load 2nd Harmonic

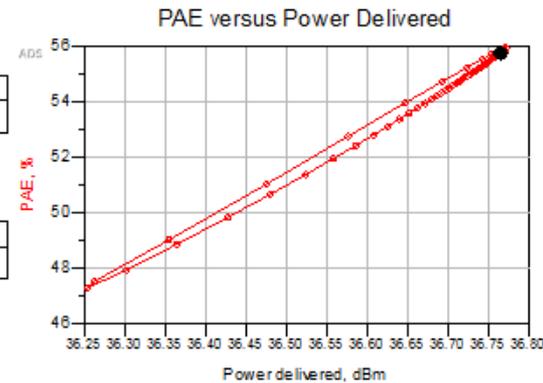


The data inside the red polygon corresponds to the load selected by the m1 marker and will only change if you move the marker to a different load.

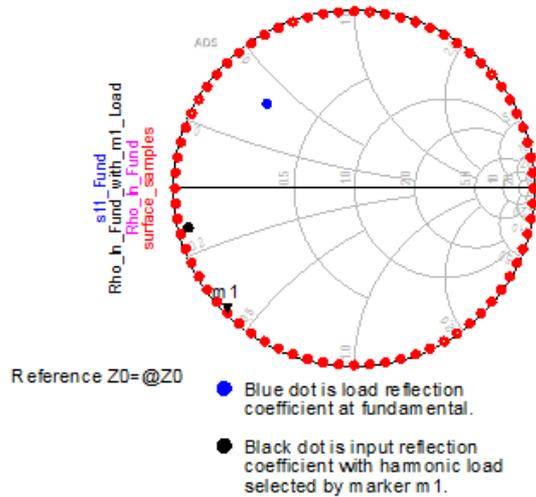


At load selected by marker m1:

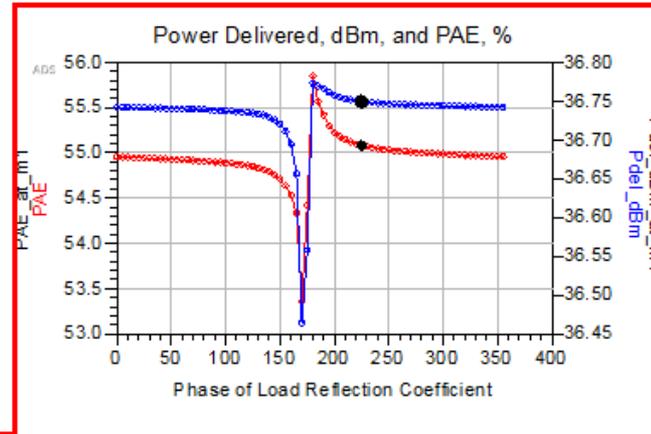
BiasCurrent_at_m1	Zload_at_m1	Rho_at_m1
0.316	1.407 + j107.210	0.990 / 50.000
PAE_at_m1		
55.741		
Z_in_at_m1	Gain_at_m1	
1.631 - j5.901	11.265	
Pdel_dBm_at_m1		
36.765		



Device Level Simulations – Load 3rd Harmonic

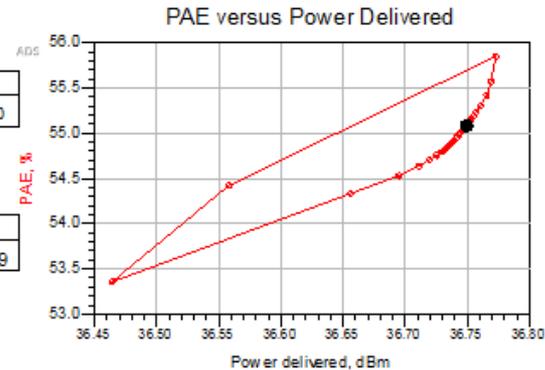


The data inside the red polygon corresponds to the load selected by the m1 marker and will only change if you move the marker to a different load.



At load selected by marker m1:

BiasCurrent_at_m1	Zload_at_m1	Rho_at_m1
0.319	0.294 - j20.710	0.990 / -135.000
PAE_at_m1	Gain_at_m1	
55.080	11.249	
Z_in_at_m1		
1.534 - j5.945		
Pdel_dBm_at_m1		
36.749		



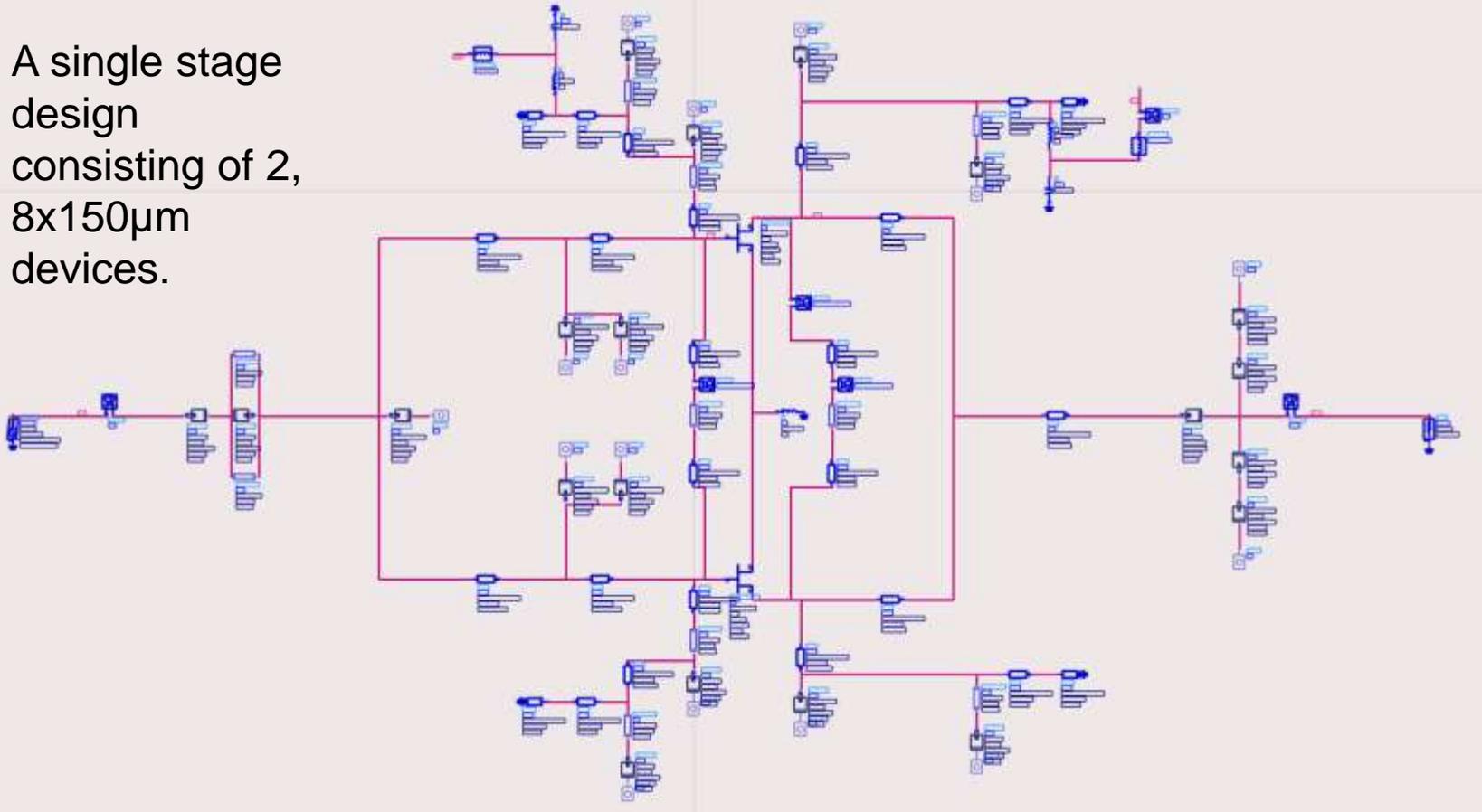
Device Level Simulations

- Unit Device Size = $8 \times 150 \mu\text{m}$
- Bias Point = $75 / 37.5 \text{ mA/mm}$ at 25V
- Fundamental and Harmonic Impedance Targets (Load / Source) @ 10GHz
 - Fundamental Load Target = $11.25\Omega + j21.67\Omega$
 - 2nd Harmonic Load Target Phase = 50° , avoid 170°
 - 3rd Harmonic Load Target Phase = -135° , avoid 170°
 - Fundamental Source Target = $2.22\Omega + j5.52\Omega$
 - 2nd Harmonic Source Target Phase = avoid $180^\circ \pm 20^\circ$
 - 3rd Harmonic Source Target Phase = avoid $-160^\circ \pm 20^\circ$
- Number of Stages = 1
- Number of Devices in Output Stage = 2
 - Total quiescent current = $180\text{mA} / 90\text{mA}$
- Expected performance:
 - $P_{\text{sat}} > 38\text{dBm}$
 - $\text{PAE} > 40\%$

Power Amplifier Schematic

(prior to layout and EM)

A single stage design consisting of 2, $8 \times 150 \mu\text{m}$ devices.

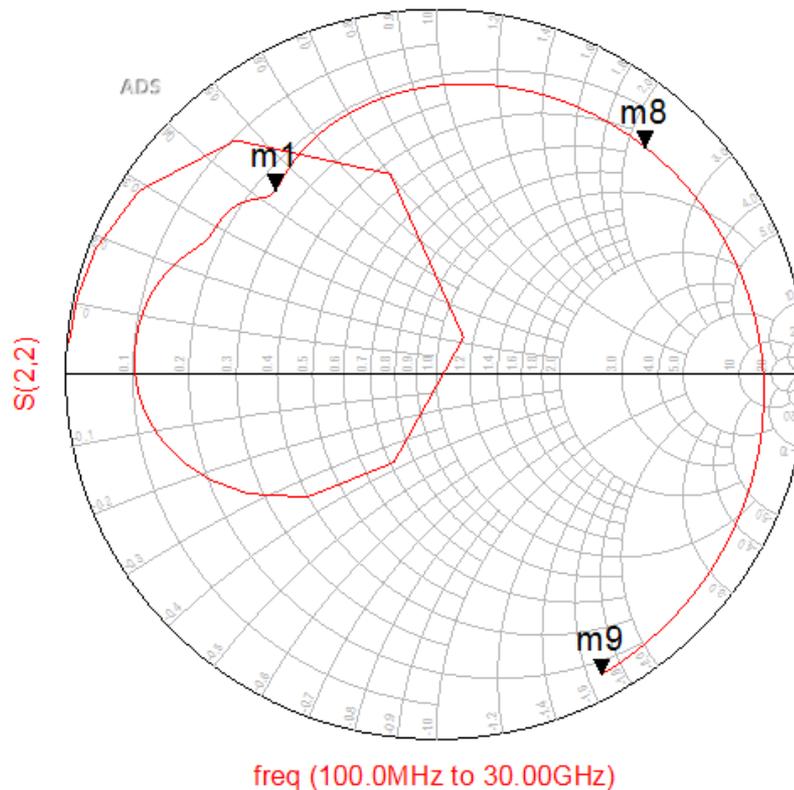


Schematic Design – Load Impedance

m1
freq=10.00GHz
S(2,2)=0.665 / 130.673
impedance = 12.081 + j21.842

m8
freq=20.00GHz
S(2,2)=0.840 / 47.833
impedance = 25.426 + j107.769

m9
freq=30.00GHz
S(2,2)=0.936 / -61.478
impedance = 6.312 - j83.727

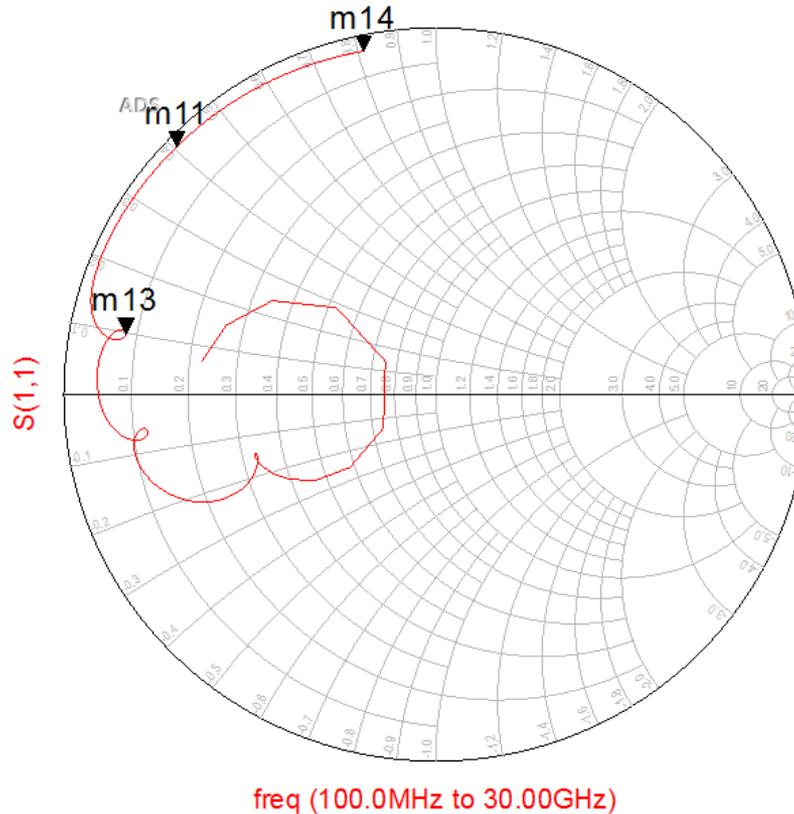


Schematic Design – Source Impedance

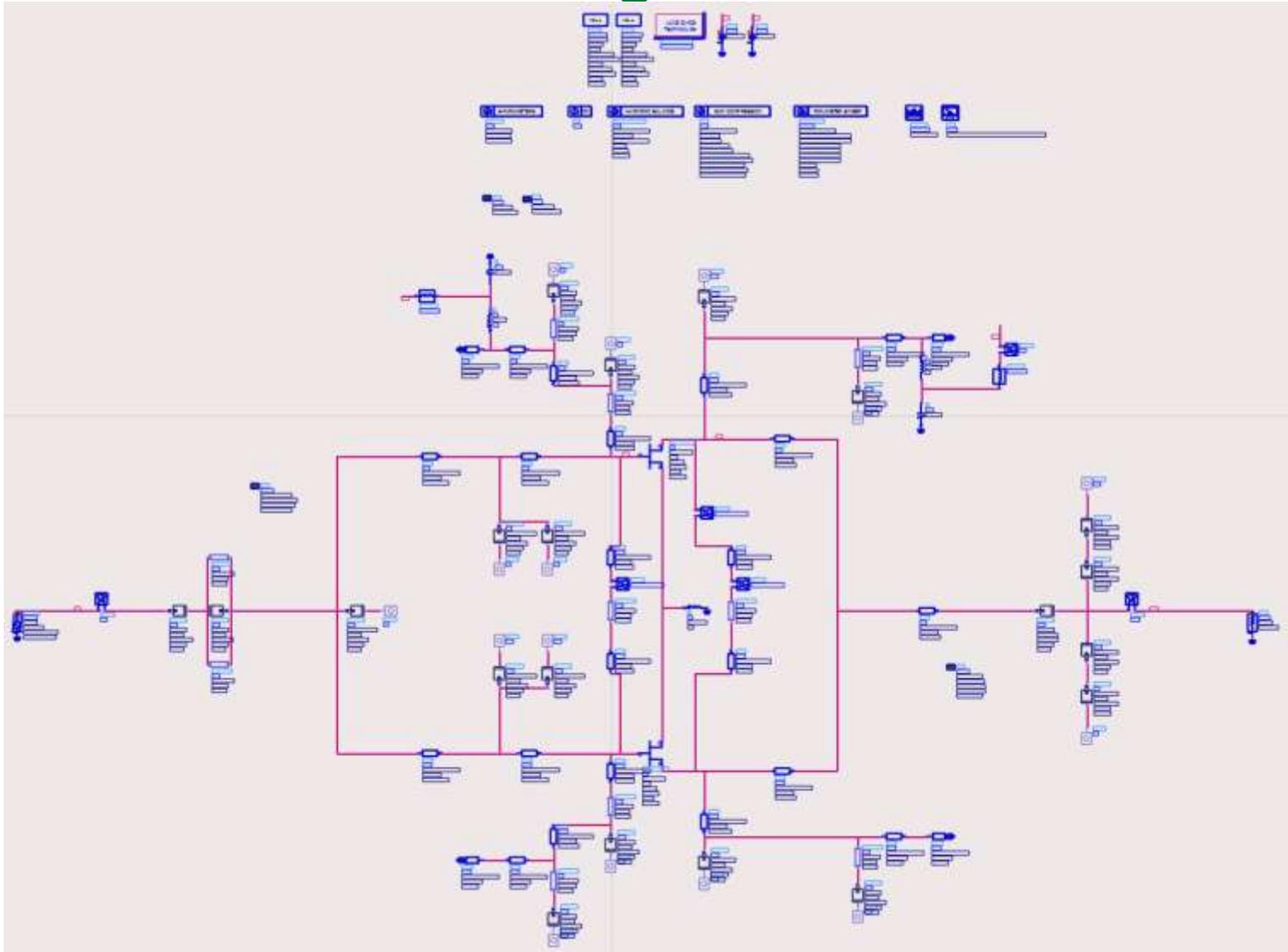
m13
freq=10.00GHz
S(1,1)=0.849 / 168.720
impedance = 4.133 + j4.904

m11
freq=20.00GHz
S(1,1)=0.971 / 135.827
impedance = 0.871 + j20.284

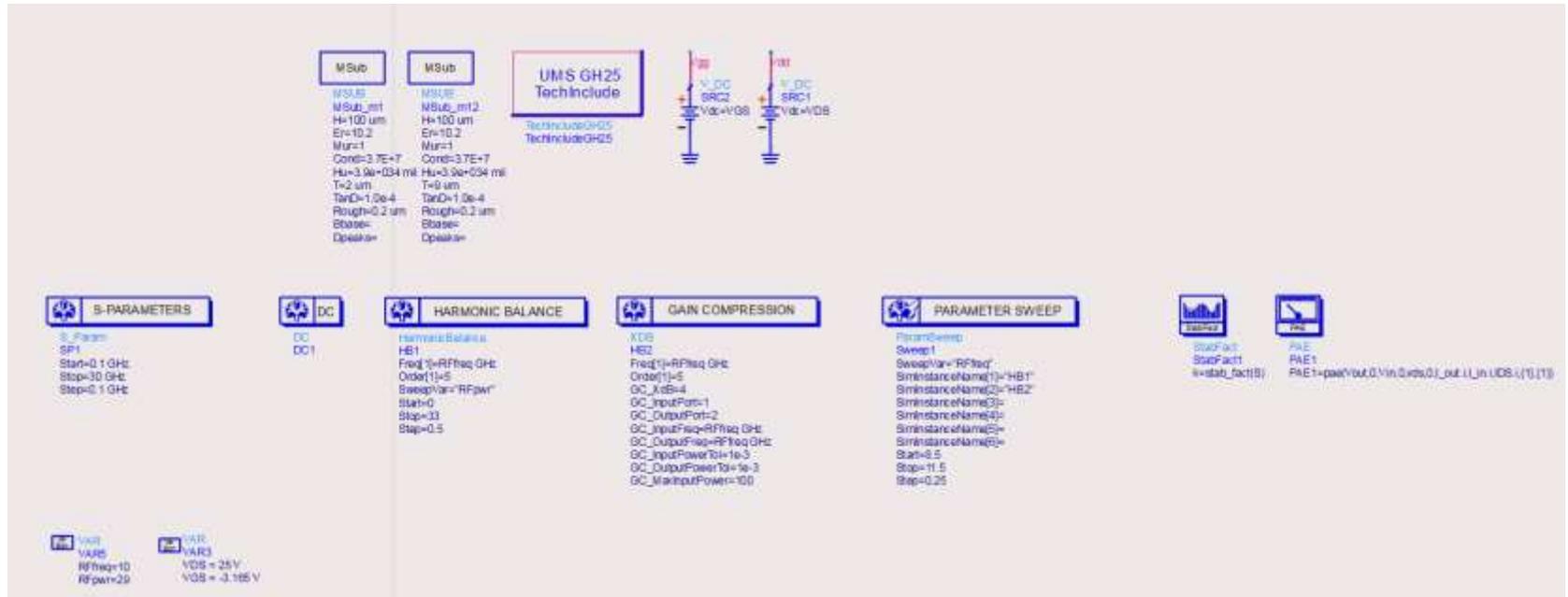
m14
freq=30.00GHz
S(1,1)=0.956 / 101.702
impedance = 1.852 + j40.671



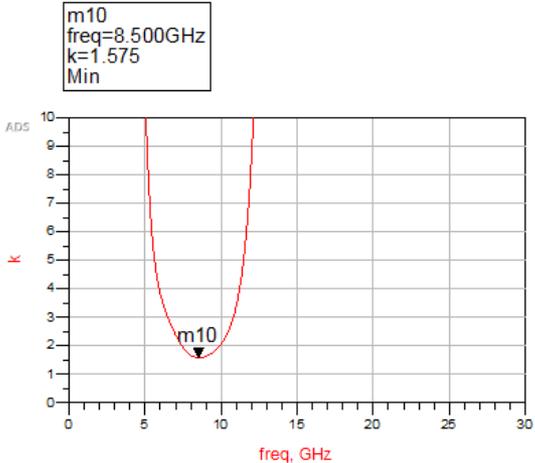
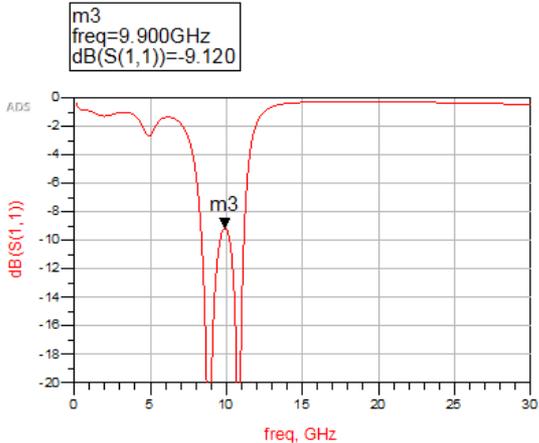
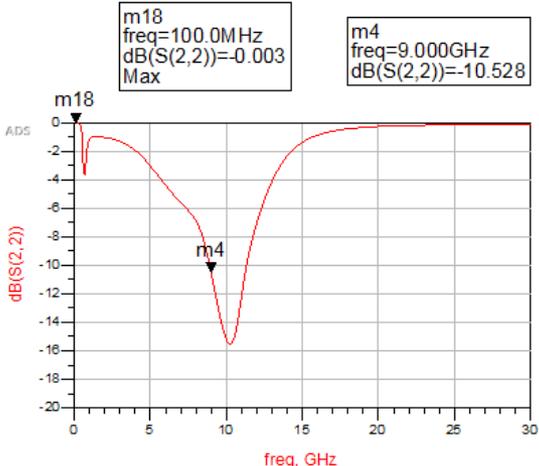
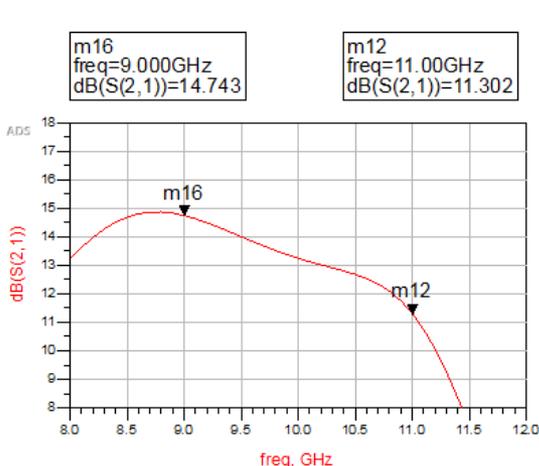
Schematic Design – Test Bench



Schematic Design – Test Bench



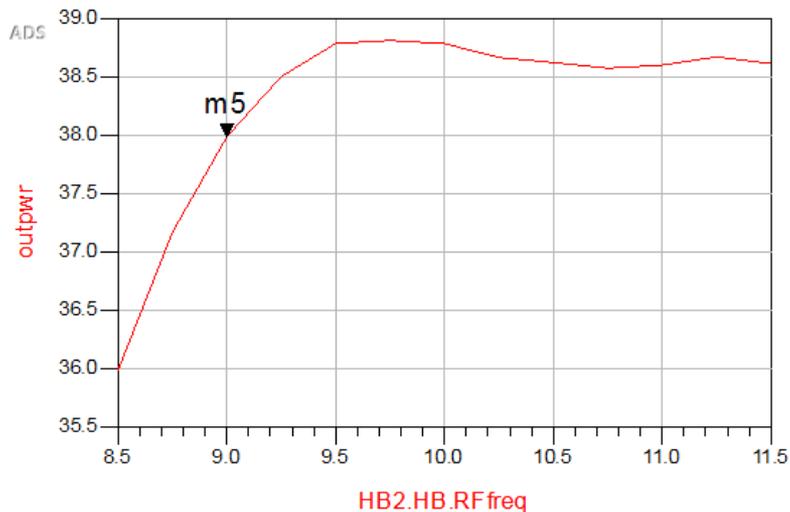
Schematic Design – Small Signal



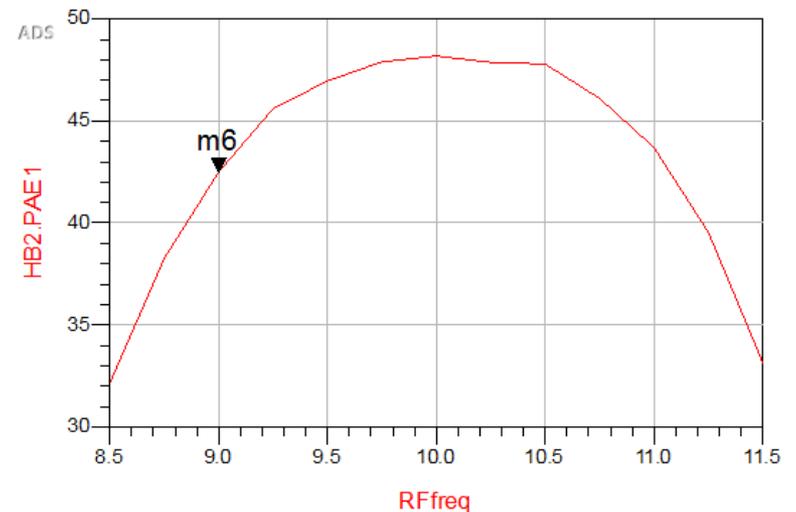
Quiescent Bias Point = 180mA at 25V

Schematic Design – Large Signal at 4dB Compression

```
m5  
indep(m5)=9.000  
plot_vs(outpwr, HB2.HB.RFfreq)=37.986  
hamindex=0
```

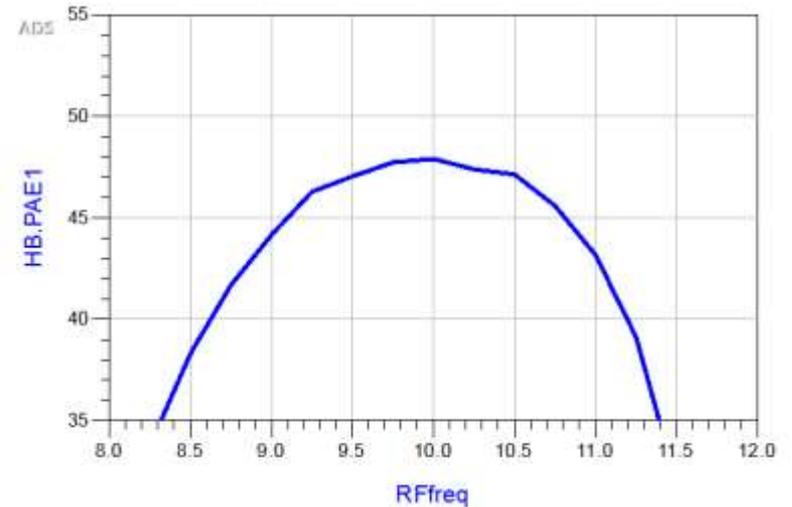
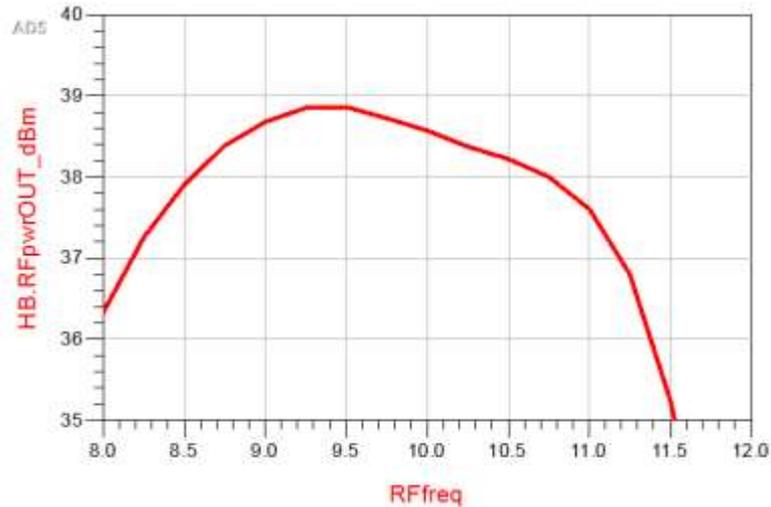


```
m6  
RFfreq=9.000  
HB2.PAE1=42.468
```



Quiescent Bias Point = 180mA at 25V

Pout and PAE for +29dBm Drive



Quiescent Bias Point = 180mA at 25V

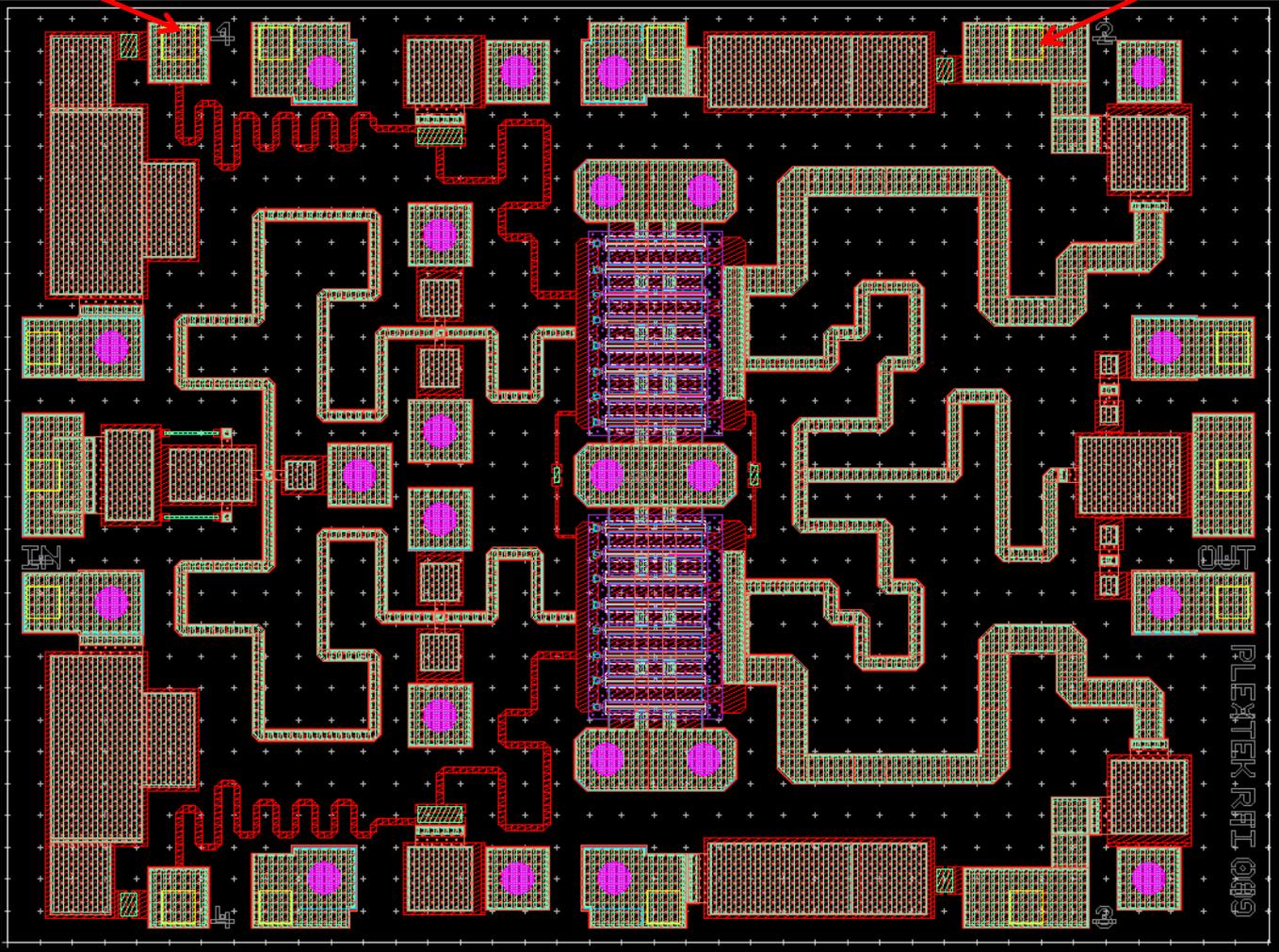
Power Amplifier Layout

Vgs

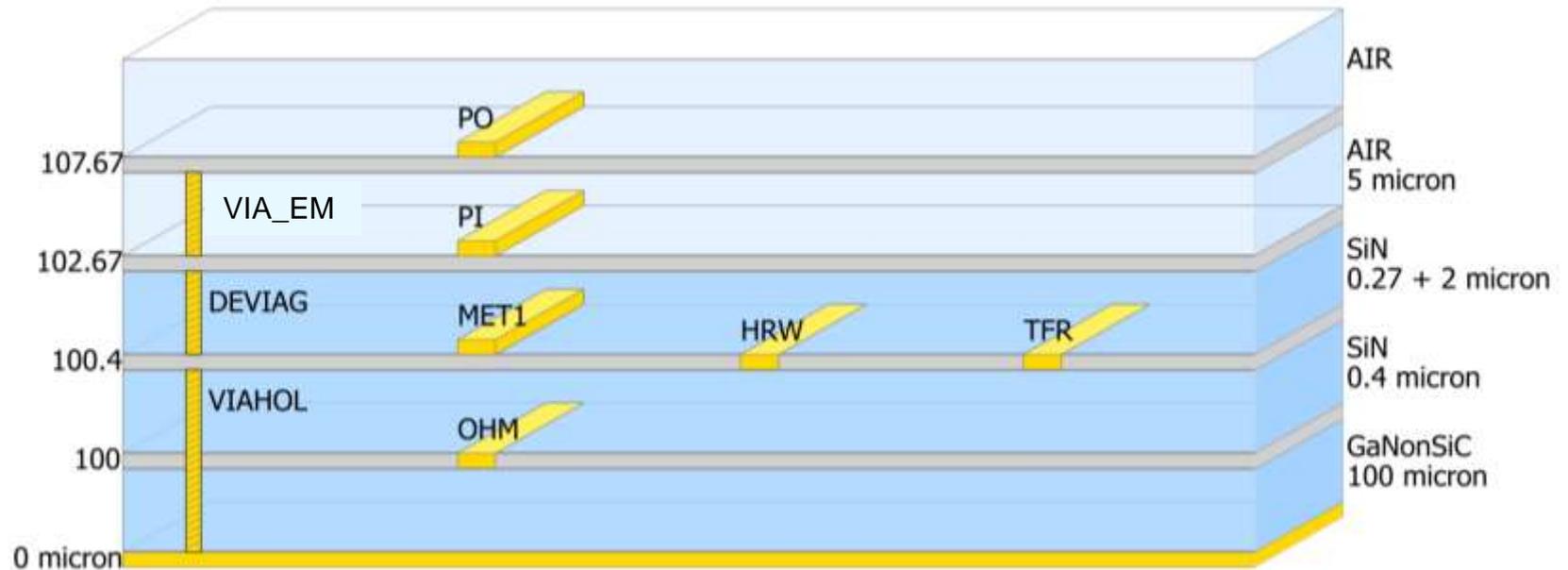
Vdd

RF I/P

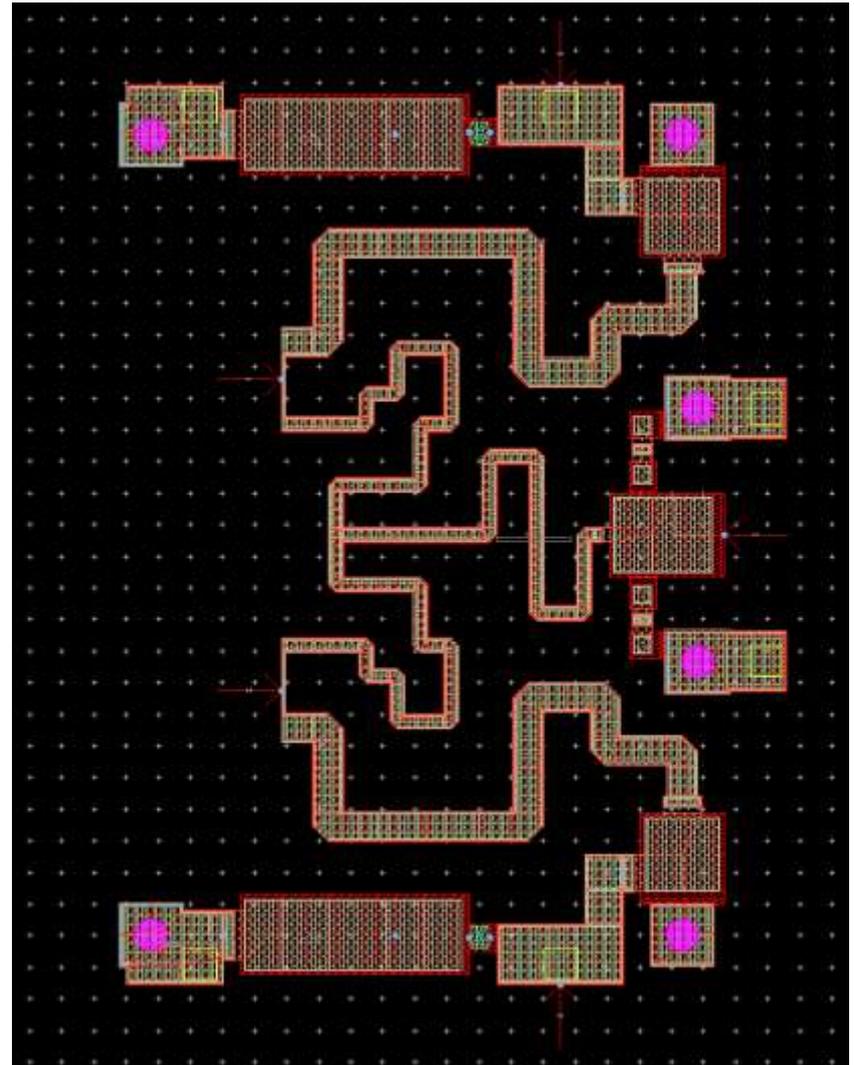
RF O/P



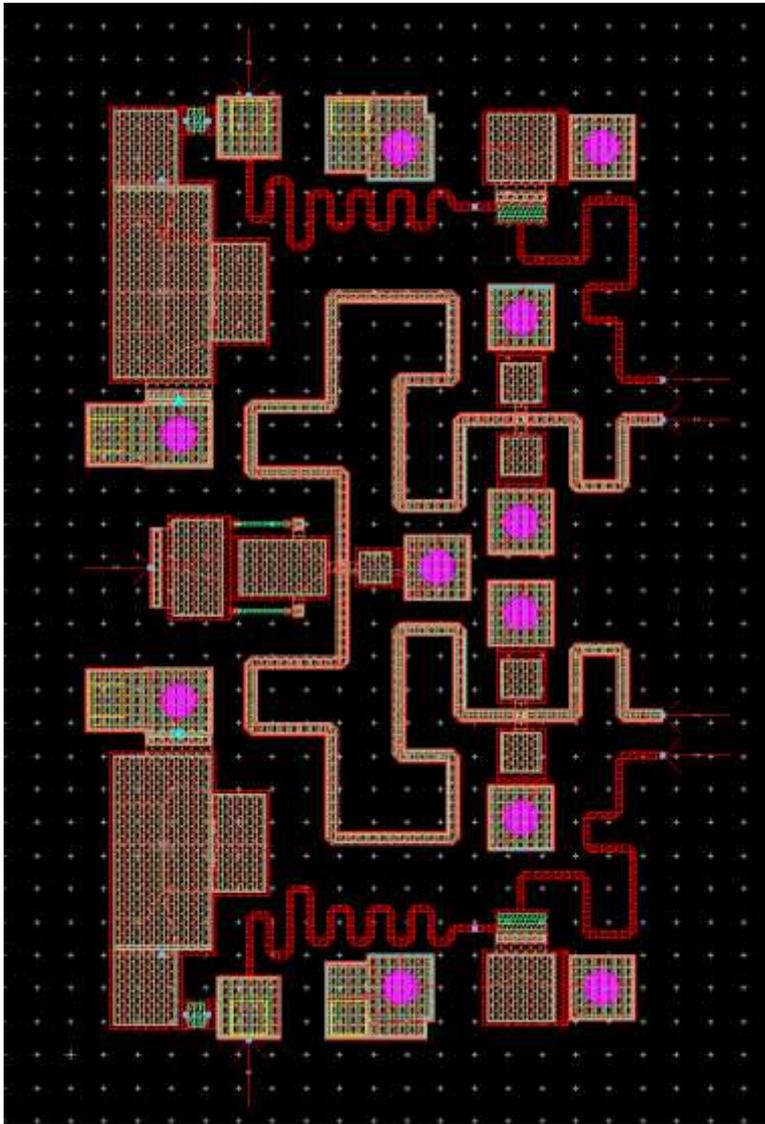
EM Simulation Stack-Up



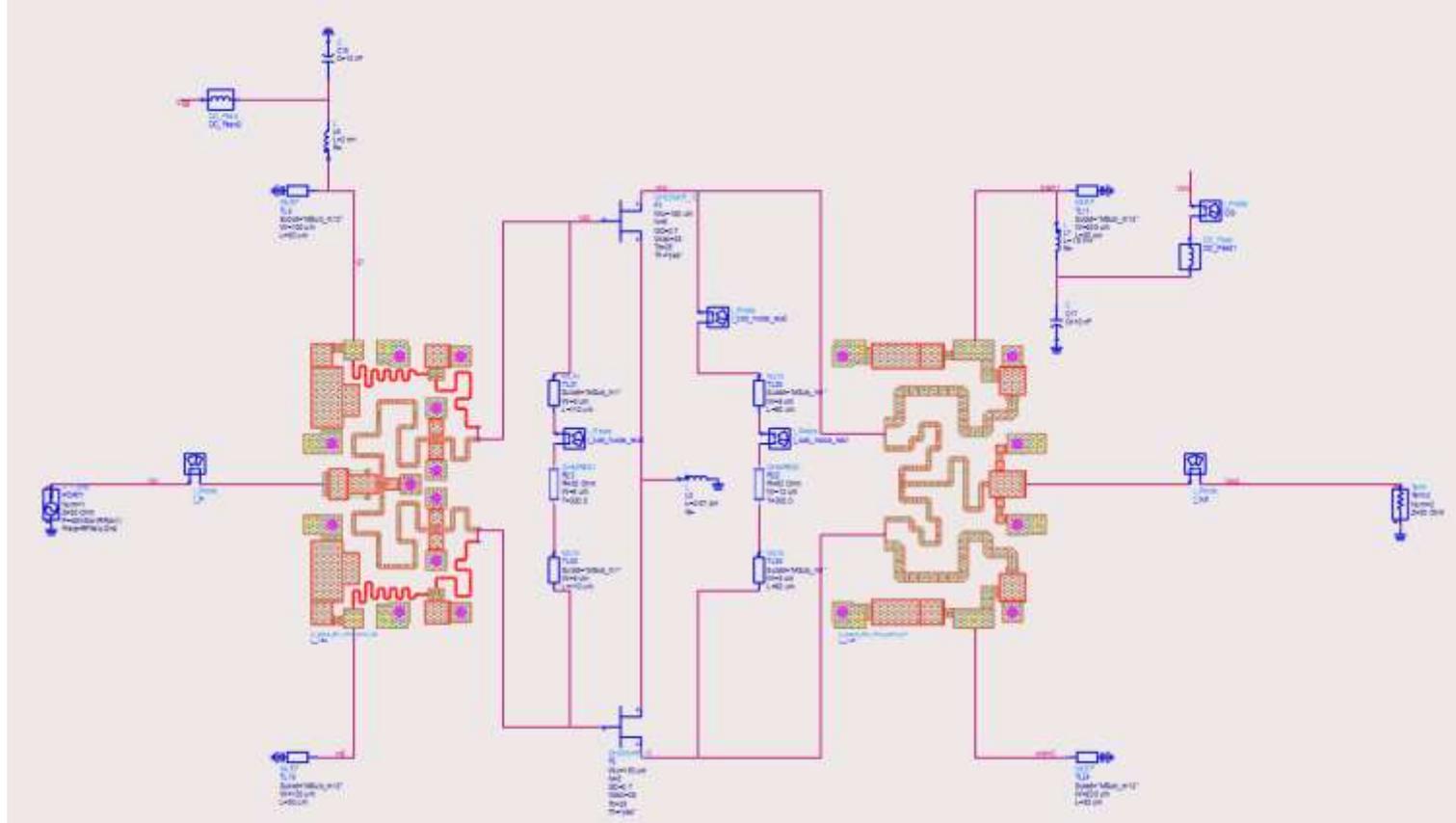
EM Simulation OPMN



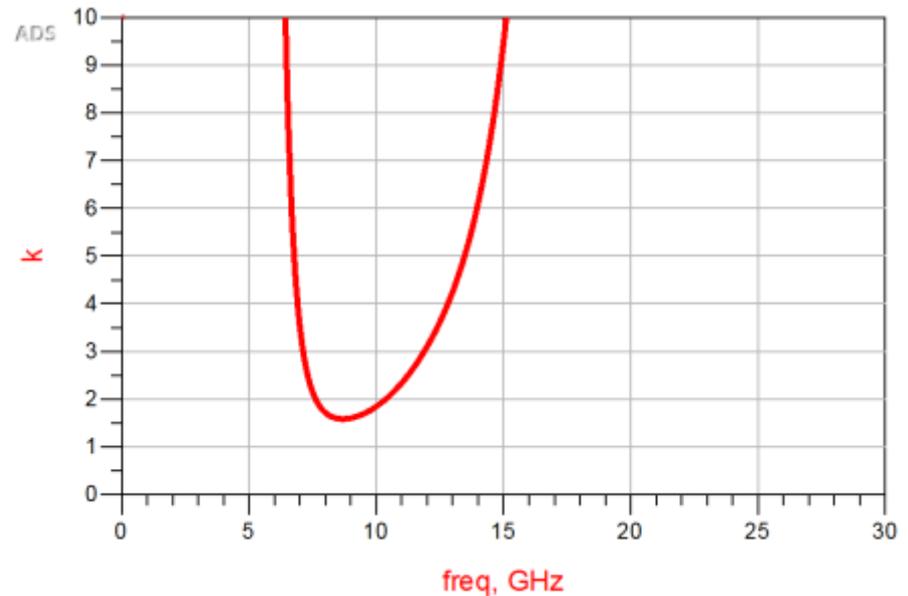
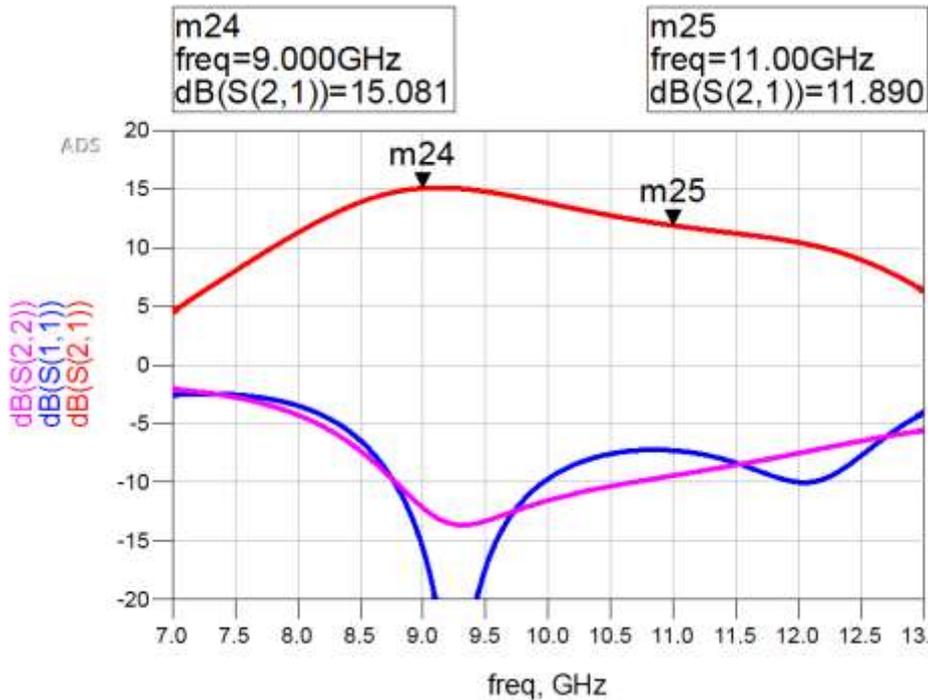
EM Simulation IPMN



Updated Schematic – Full EM Simulation

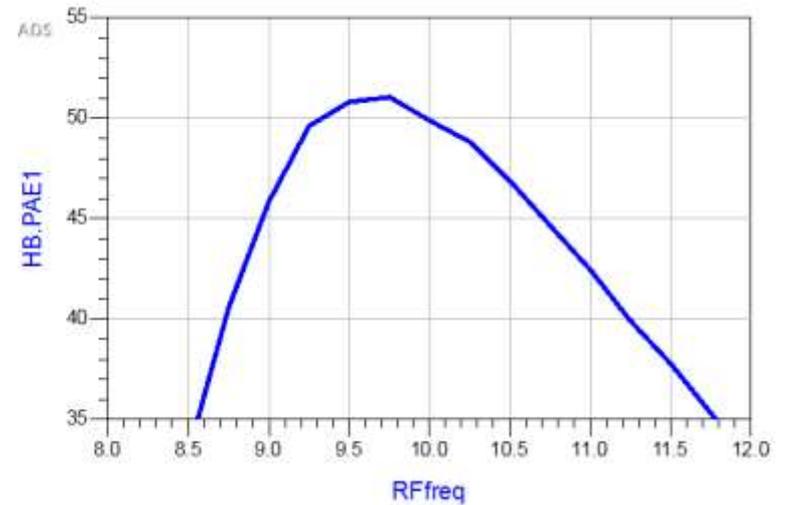
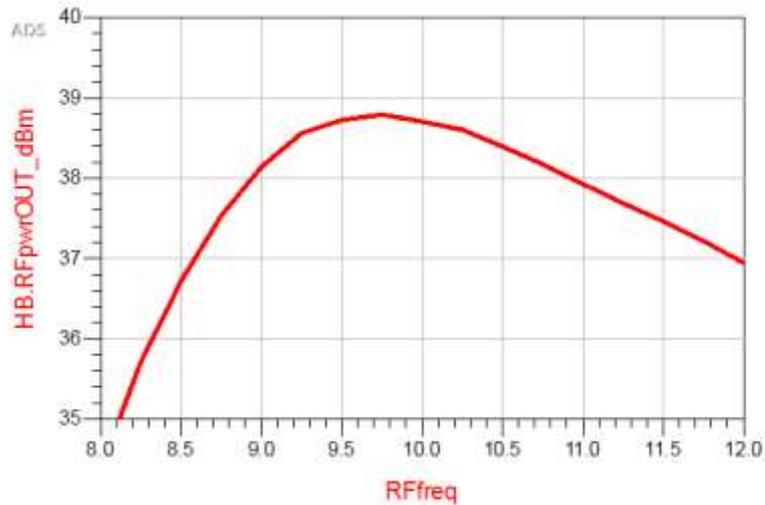


Power Amplifier Simulated Performance Including Full EM – Small Signal



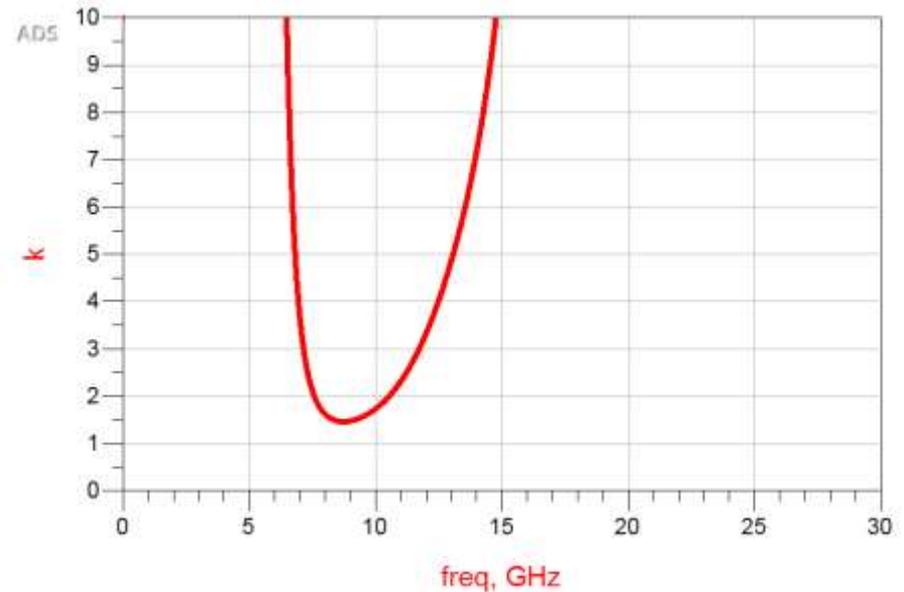
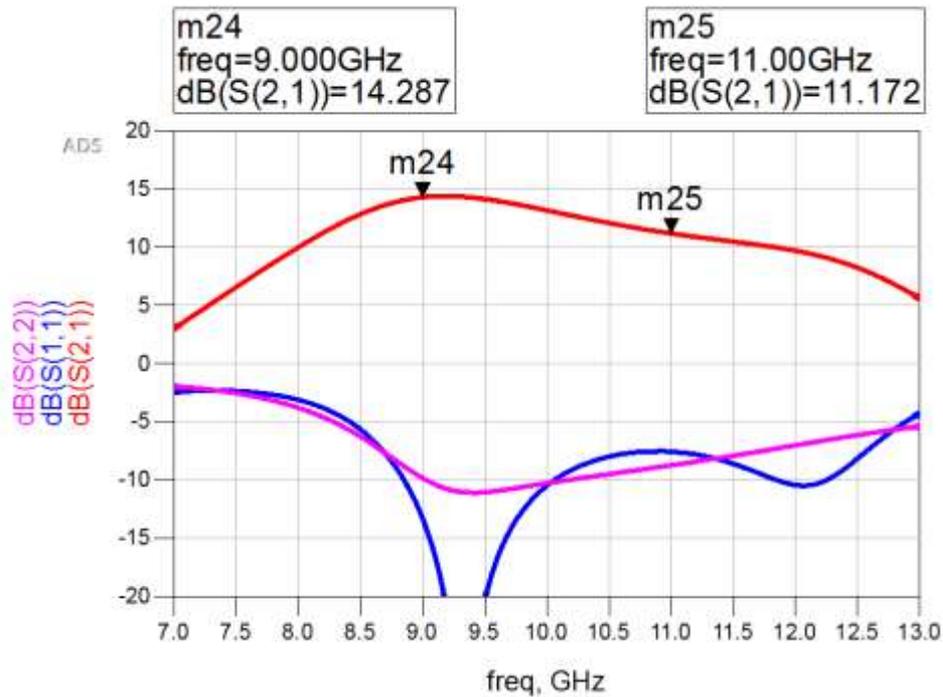
Quiescent Bias Point = 180mA at 25V

Pout and PAE for +29dBm Drive Full EM Simulation



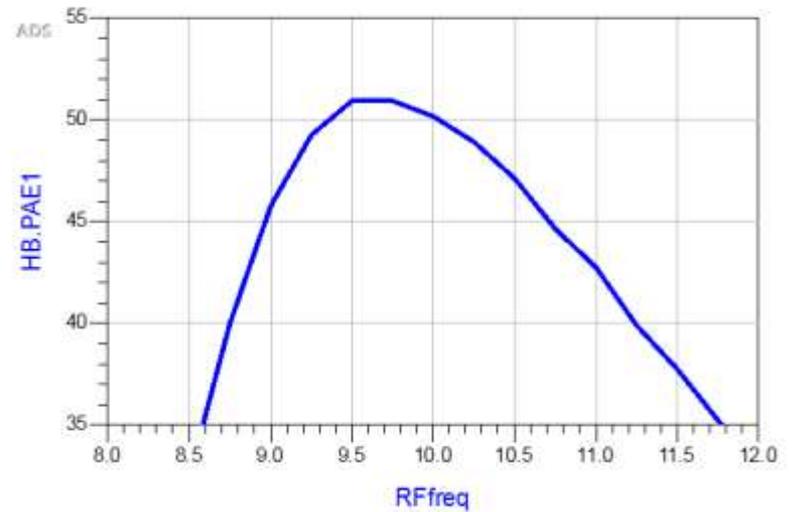
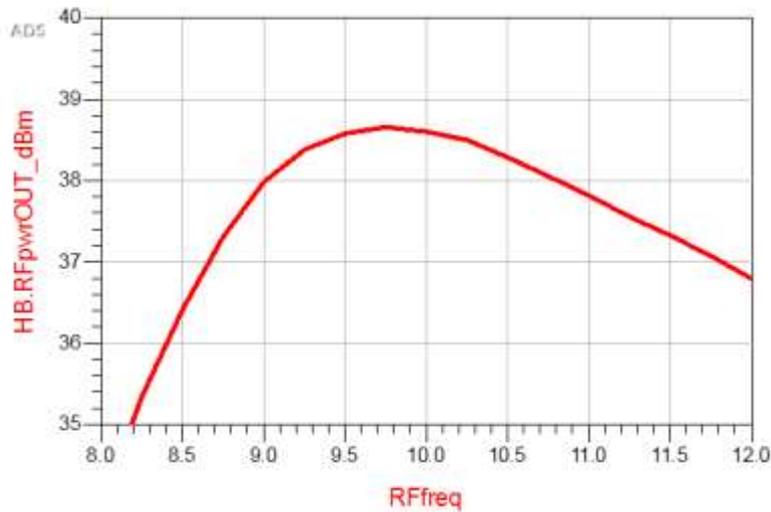
Quiescent Bias Point = 180mA at 25V

Power Amplifier Simulated Performance Including Full EM – Small Signal



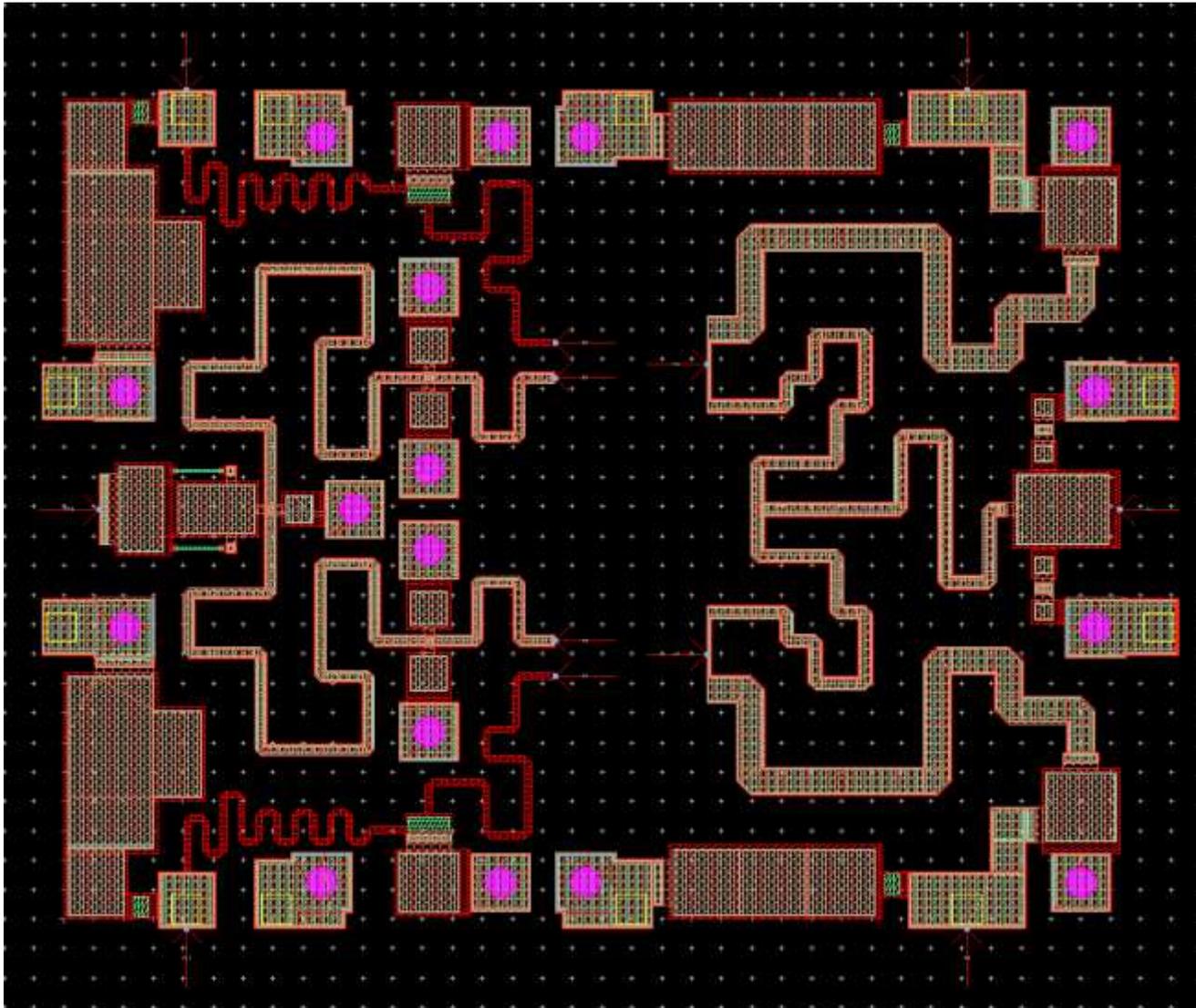
Quiescent Bias Point = 90mA at 25V

Pout and PAE for +29dBm Drive Full EM Simulation

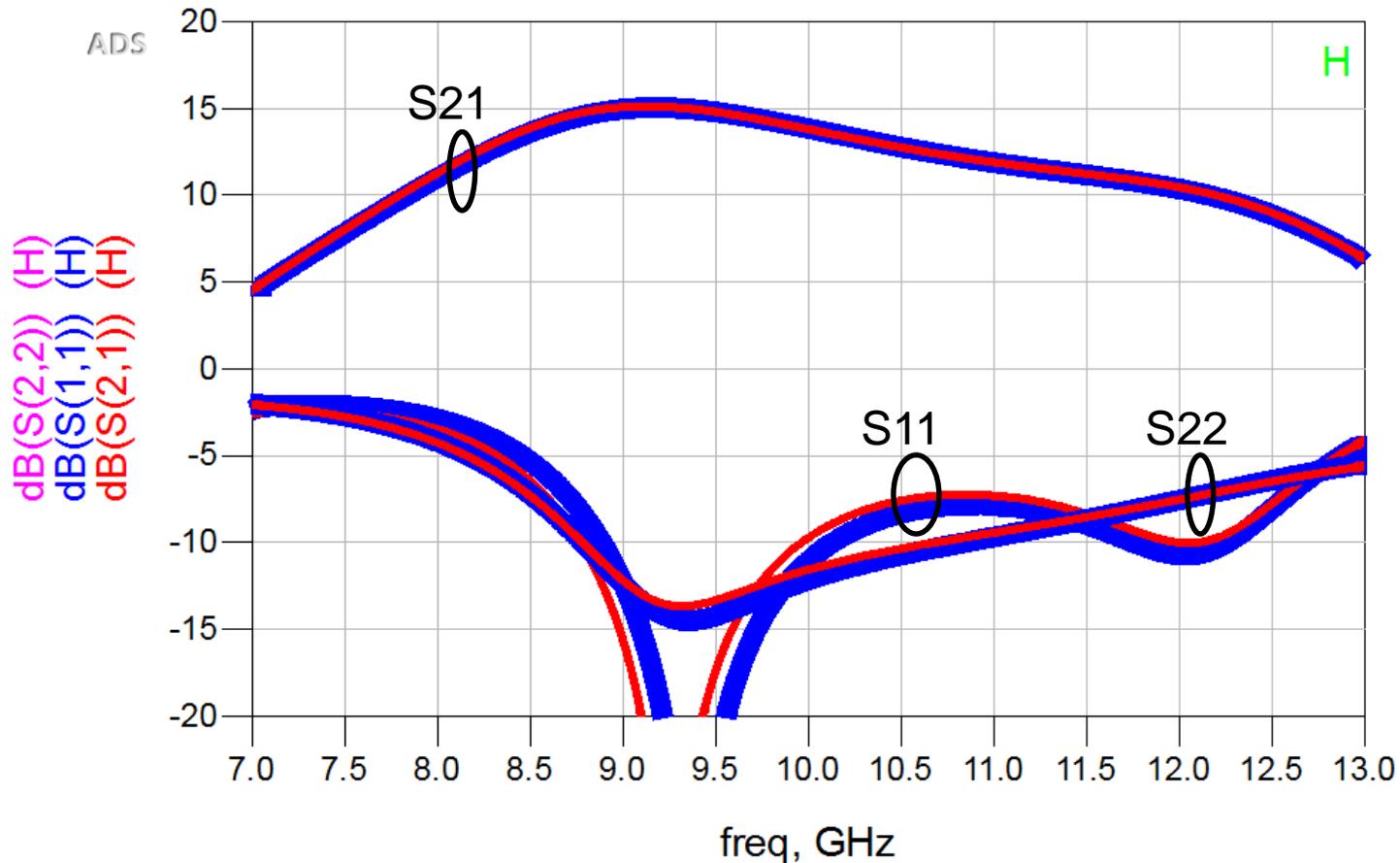


Quiescent Bias Point = 90mA at 25V

EM Simulation IPMN and OPMN Together

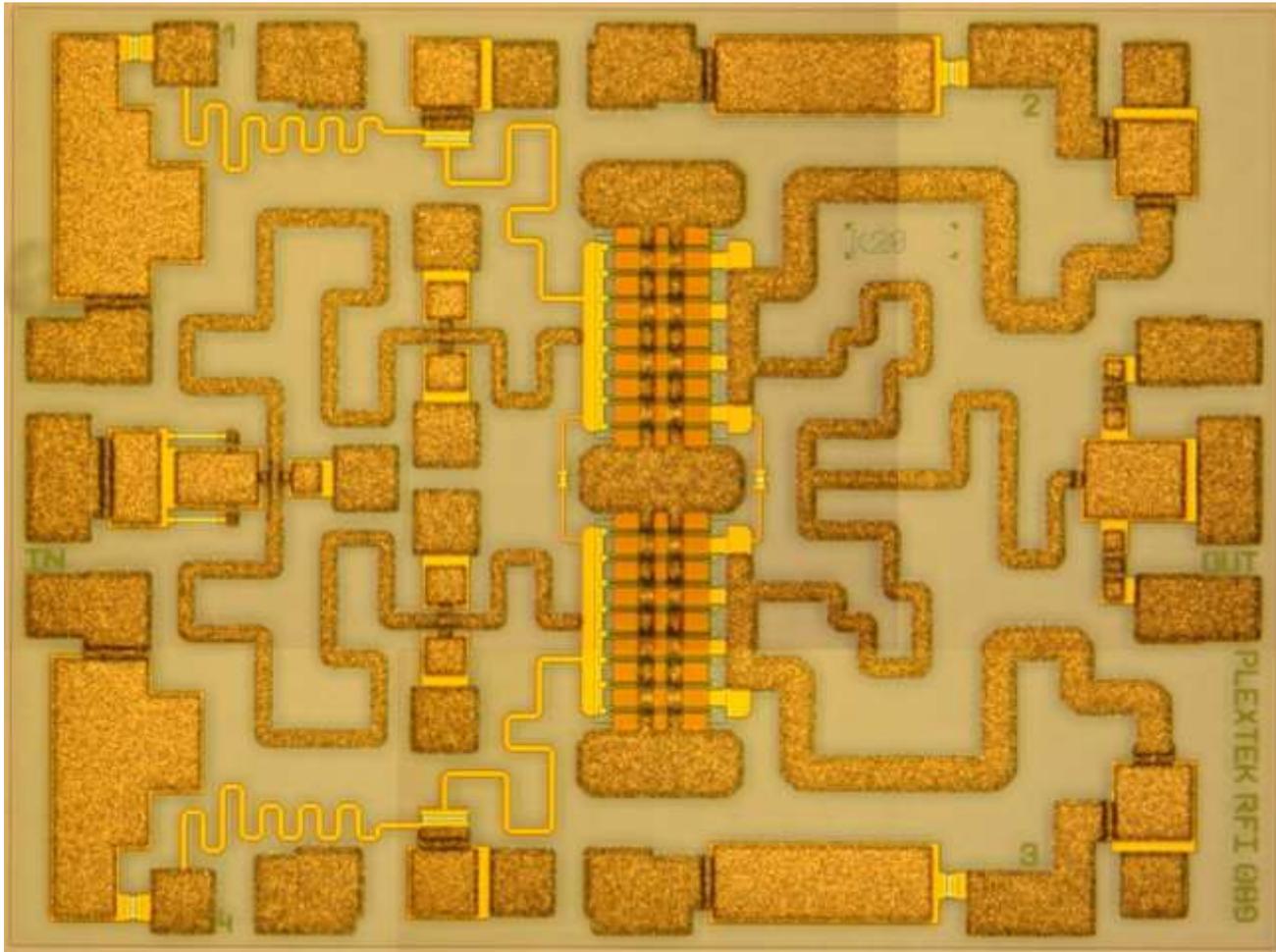


Comparing effect of EM simulating IPMN and OPMN together and separately



Quiescent Bias Point = 180mA at 25V

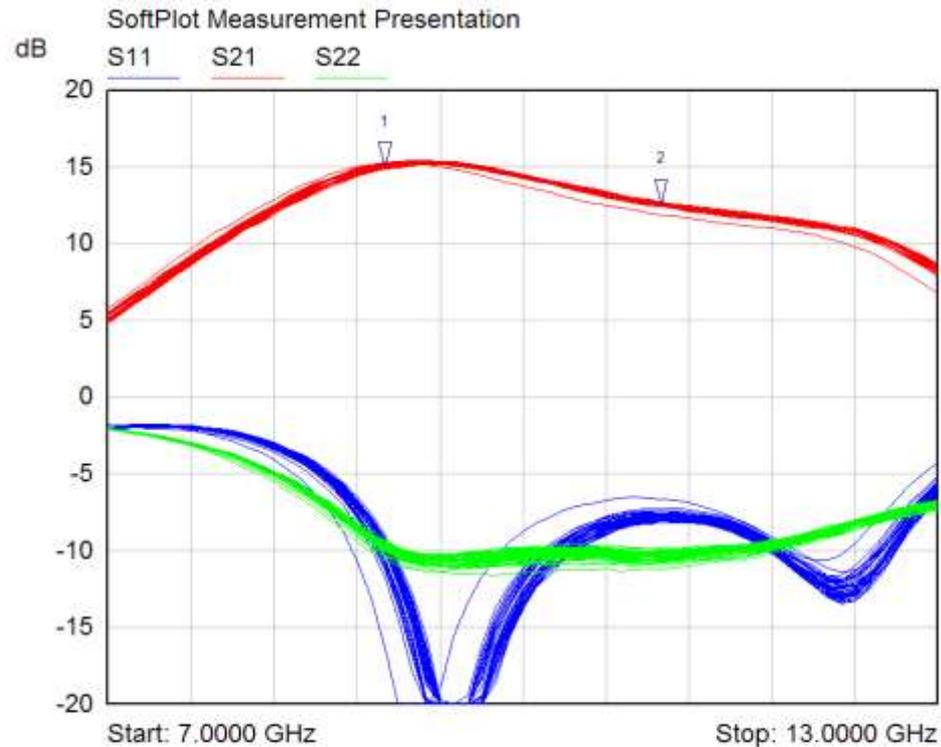
Photograph of the GaN PA MMIC



Die size: 1.5mm x 2mm (around 2,300 die per 4" diameter wafer)

Power Amplifier Measured S-Parameters

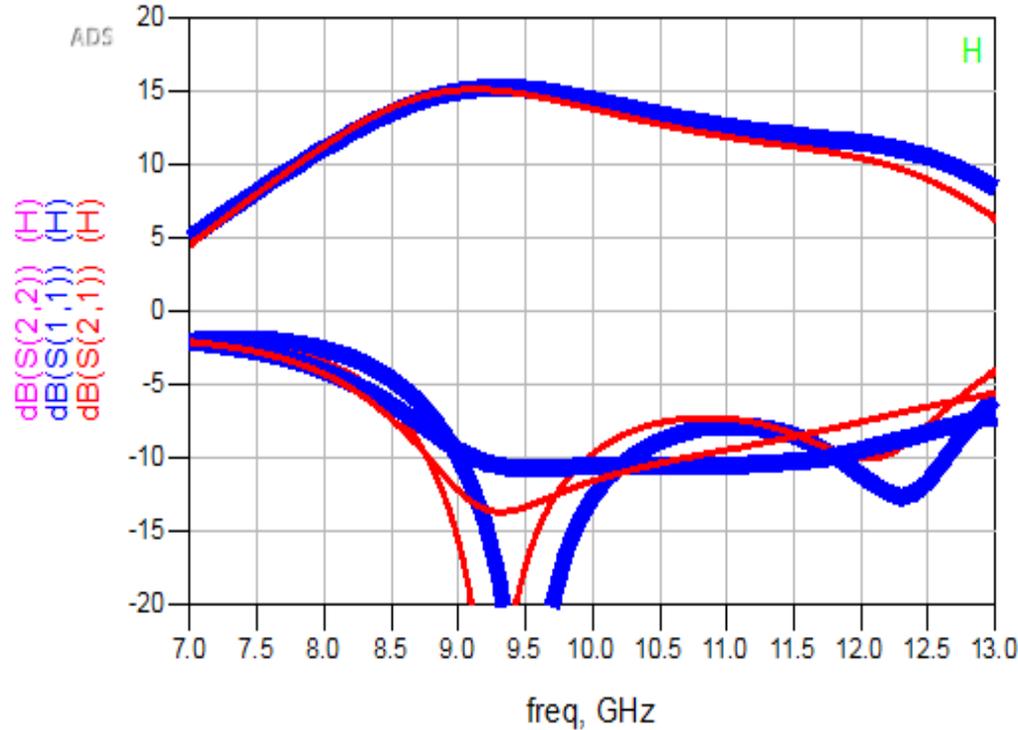
40 devices



Mkr	Trace	X-Axis	Value	Notes
1 ▽	S21	9.0000 GHz	15.00 dB	
2 ▽	S21	11.0000 GHz	12.60 dB	

Quiescent Bias Point = 180mA at 25V

Power Amplifier Measured Vs Modelled S-Parameters

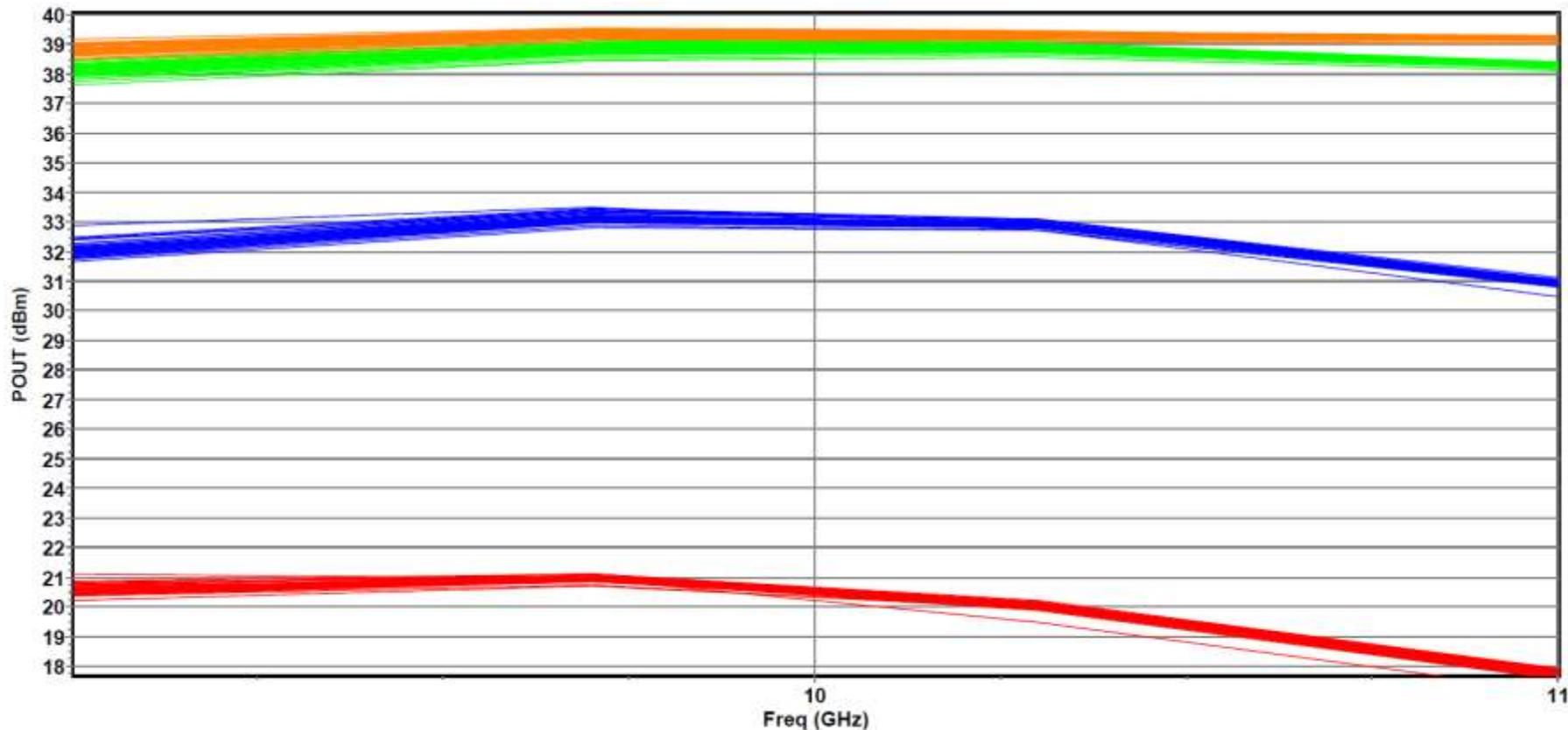


Measured
in blue
Modelled in
red

Quiescent Bias Point = 180mA at 25V

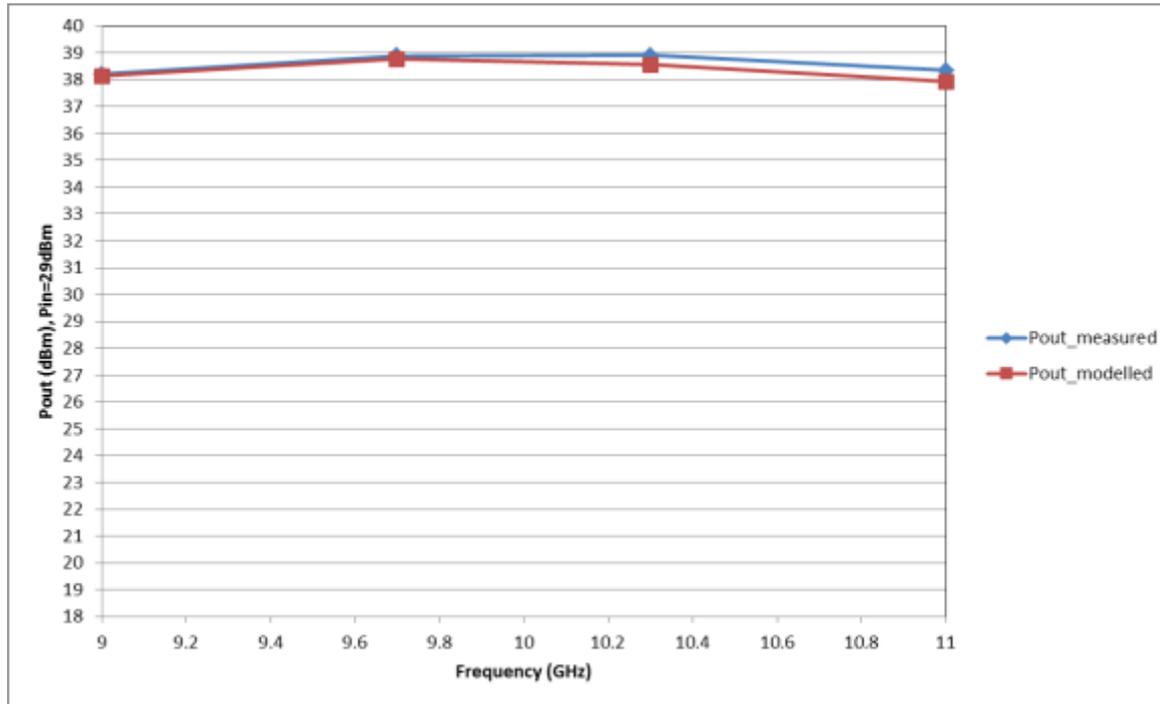
RFOW Measured Output Power

25 μ s and 10% Duty Cycle, input powers: **5dBm**, **19dBm**,
29dBm, **32dBm**



Quiescent Bias Point = 180mA at 25V

Power Amplifier Measured Vs Modelled Pout for +29dBm input power

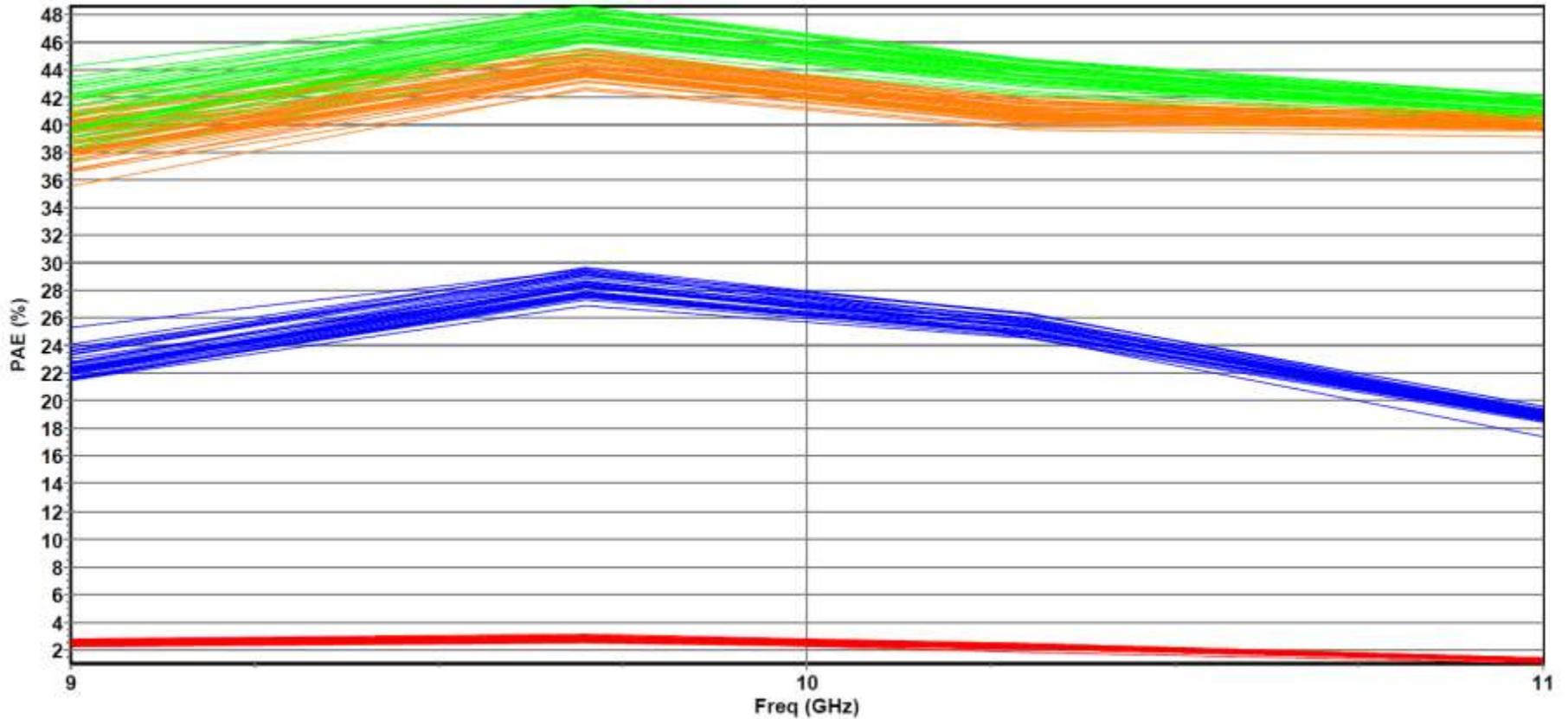


Measured in
blue
Modelled in
red

Quiescent Bias Point = 180mA at 25V

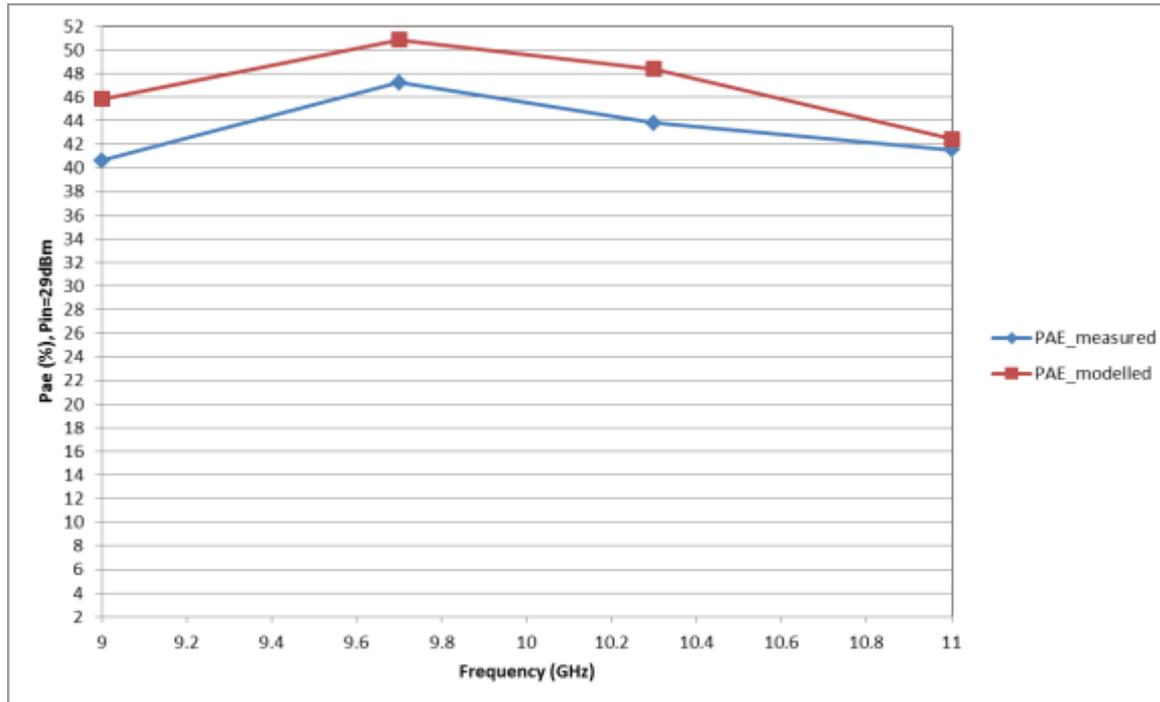
RFOW Measured PAE

25 μ s and 10% Duty Cycle, input powers: 5dBm, 19dBm, 29dBm, 32dBm



Quiescent Bias Point = 180mA at 25V

Power Amplifier Measured Vs Modelled PAE for +29dBm input power

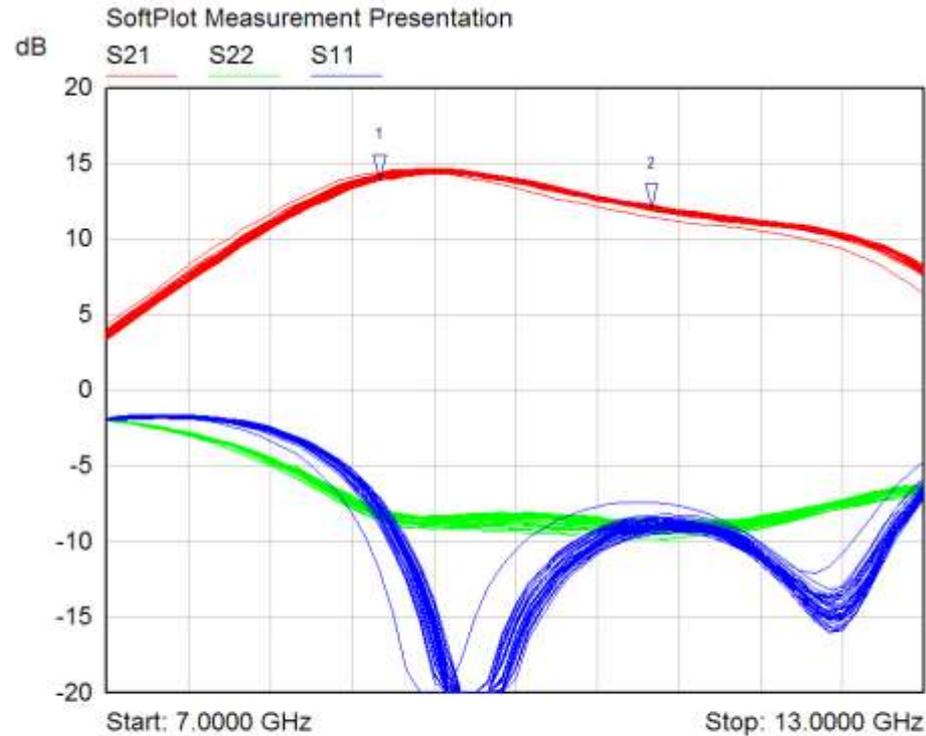


Measured in
blue
Modelled in
red

Quiescent Bias Point = 180mA at 25V

Power Amplifier Measured S-Parameters

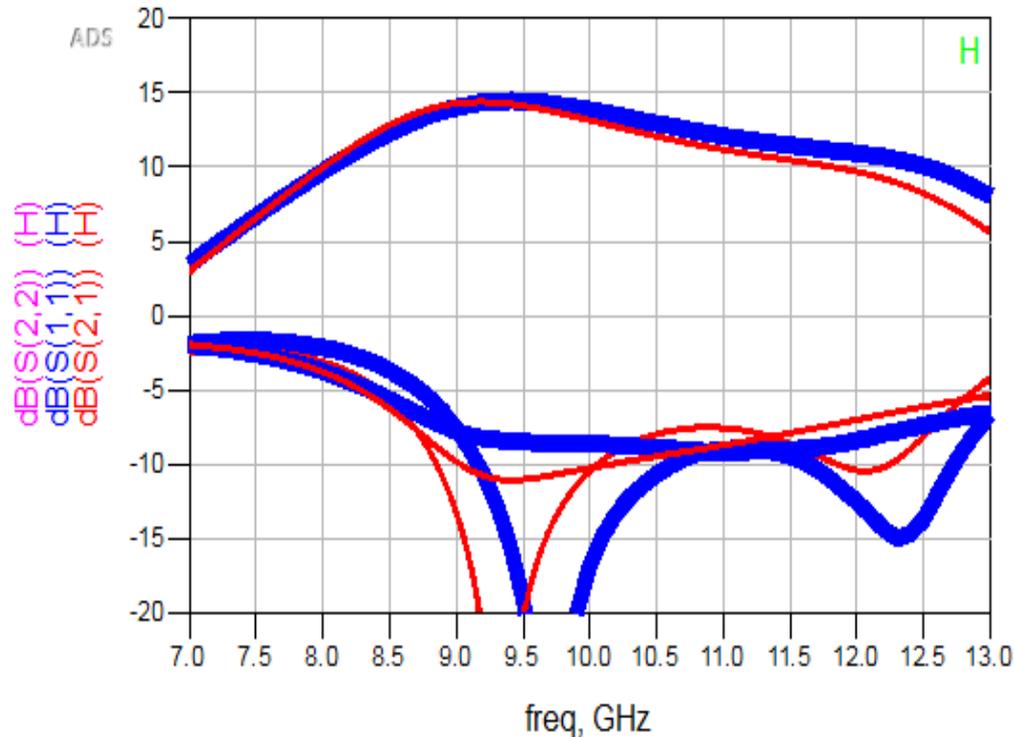
40 devices



Mkr	Trace	X-Axis	Value	Notes
1 ▽	S21	9.0000 GHz	14.00 dB	
2 ▽	S21	11.0000 GHz	12.09 dB	

Quiescent Bias Point = 90mA at 25V

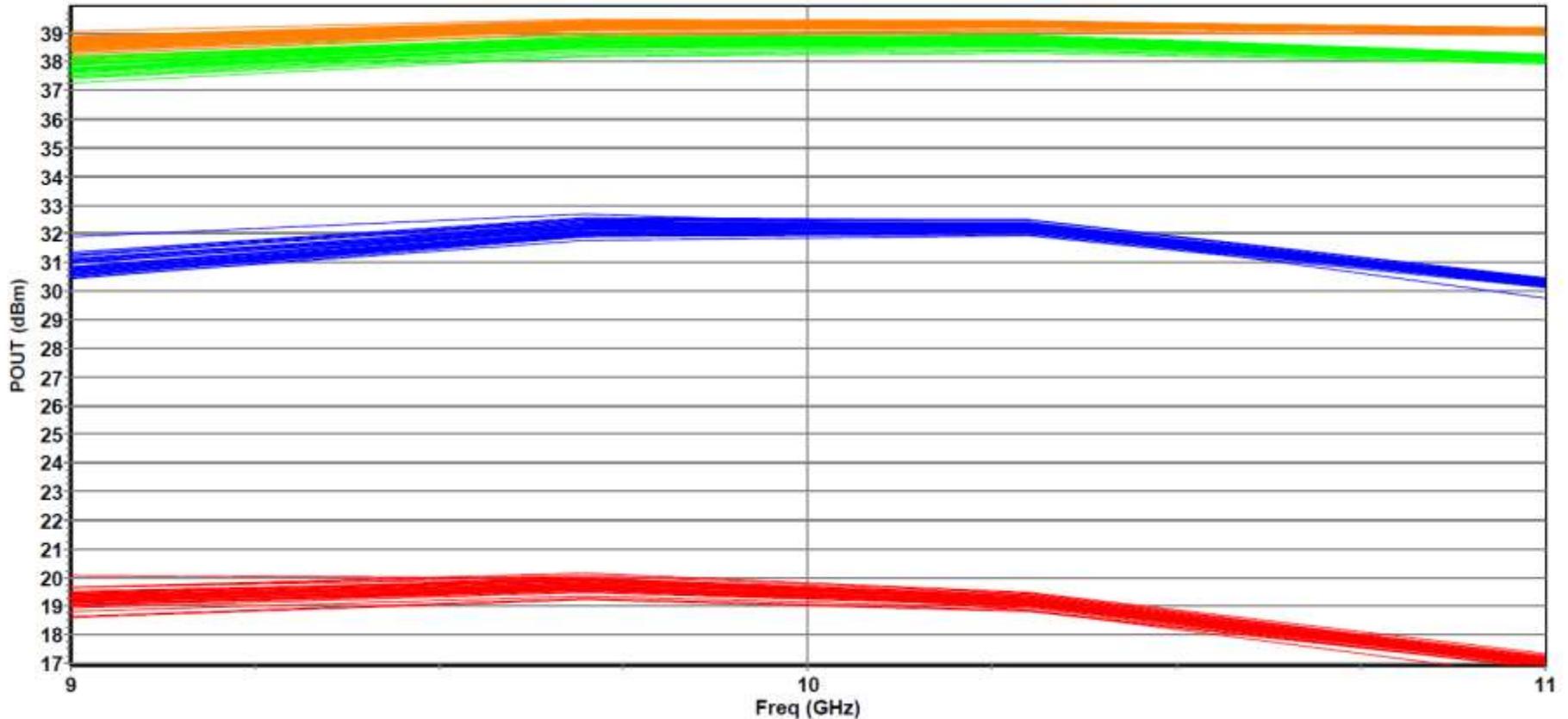
Power Amplifier Measured Vs Modelled S-Parameters



Quiescent Bias Point = 90mA at 25V

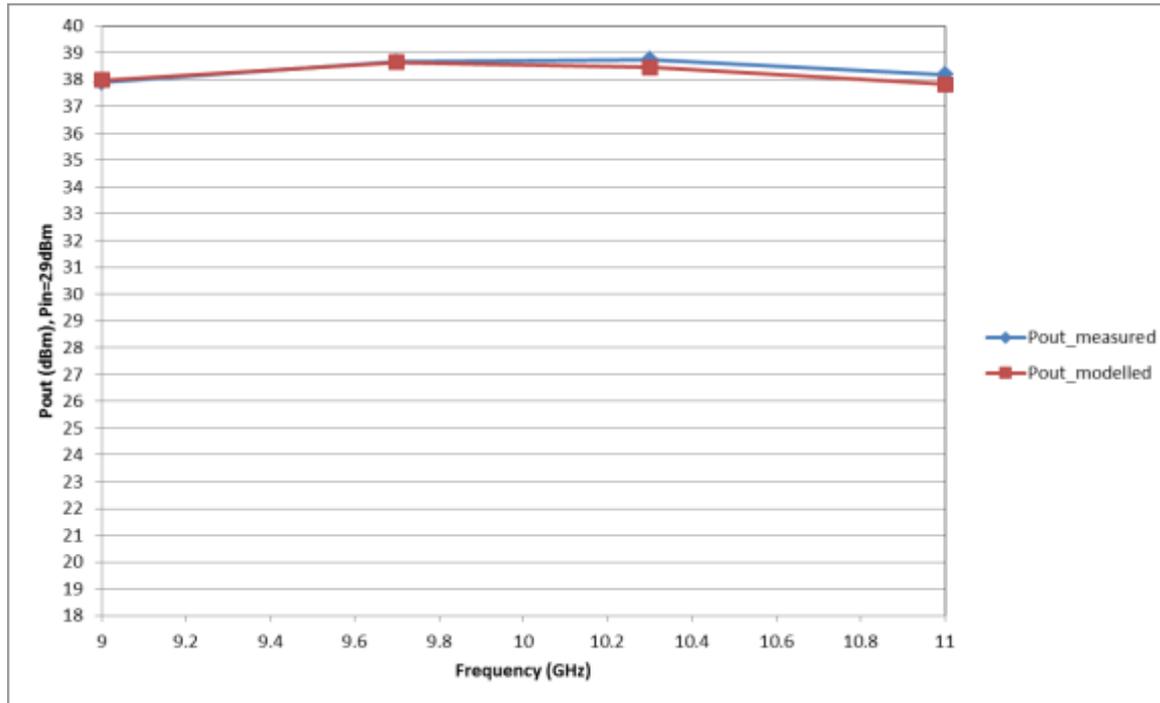
RFOW Measured Output Power

25 μ s and 10% Duty Cycle, input powers: **5dBm**, **19dBm**,
29dBm, **32dBm**



Quiescent Bias Point = 90mA at 25V

Power Amplifier Measured Vs Modelled Pout for +29dBm input power

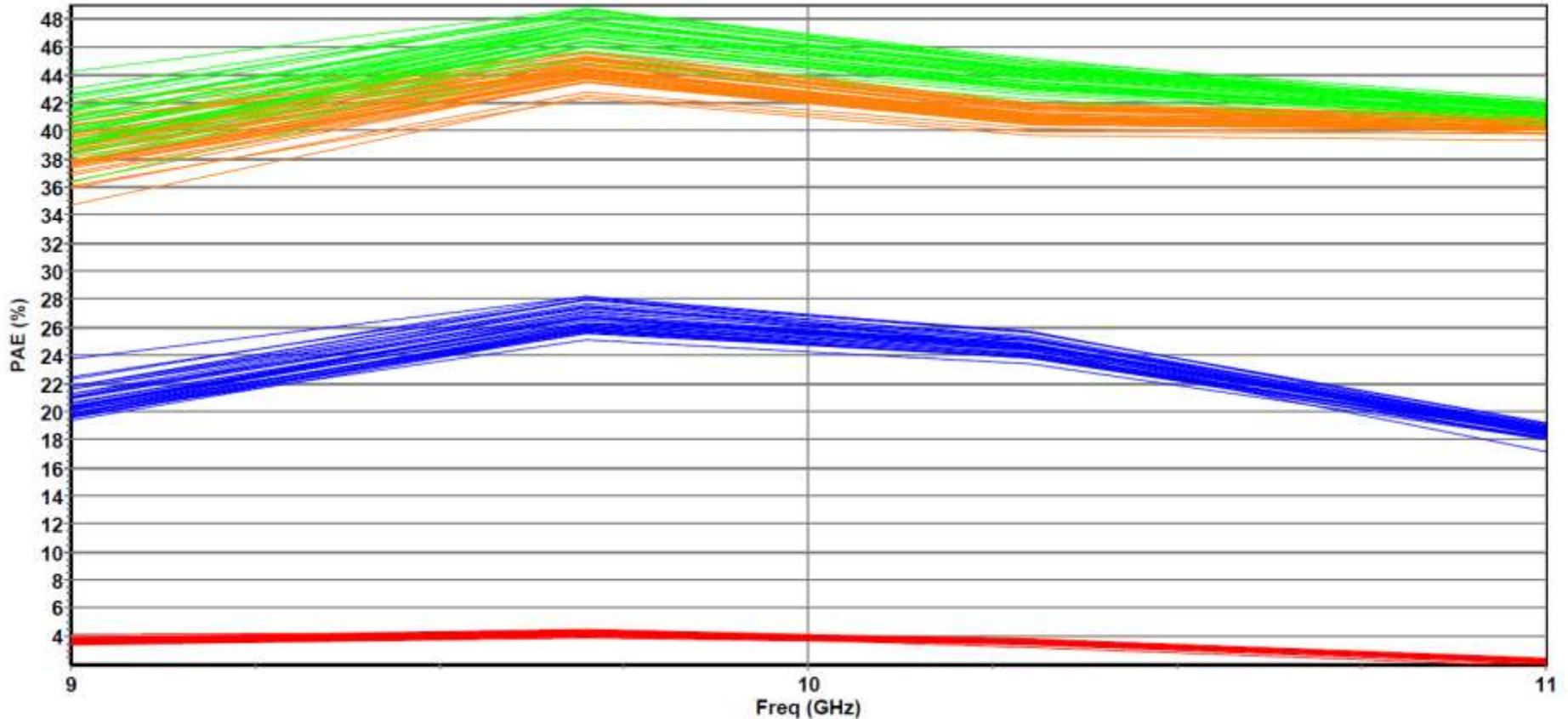


Measured in
blue
Modelled in
red

Quiescent Bias Point = 90mA at 25V

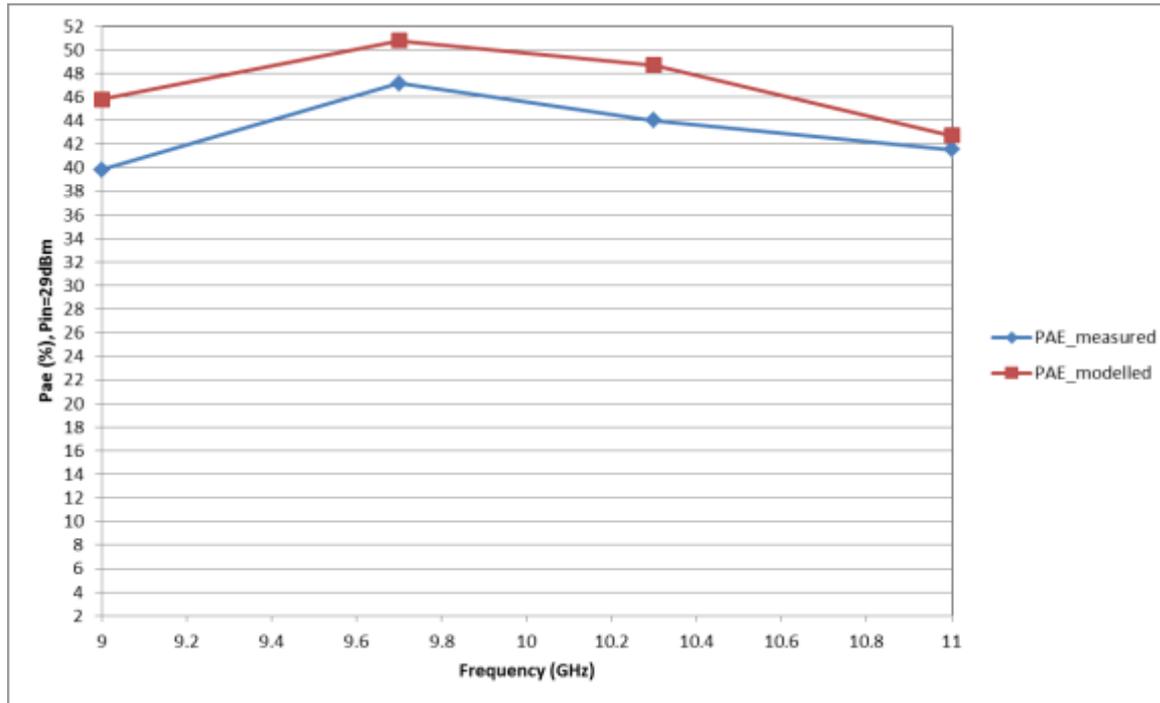
RFOW Measured PAE

25 μ s and 10% Duty Cycle, input powers: 5dBm, 19dBm, 29dBm, 32dBm



Quiescent Bias Point = 90mA at 25V

Power Amplifier Measured Vs Modelled PAE for +29dBm input power

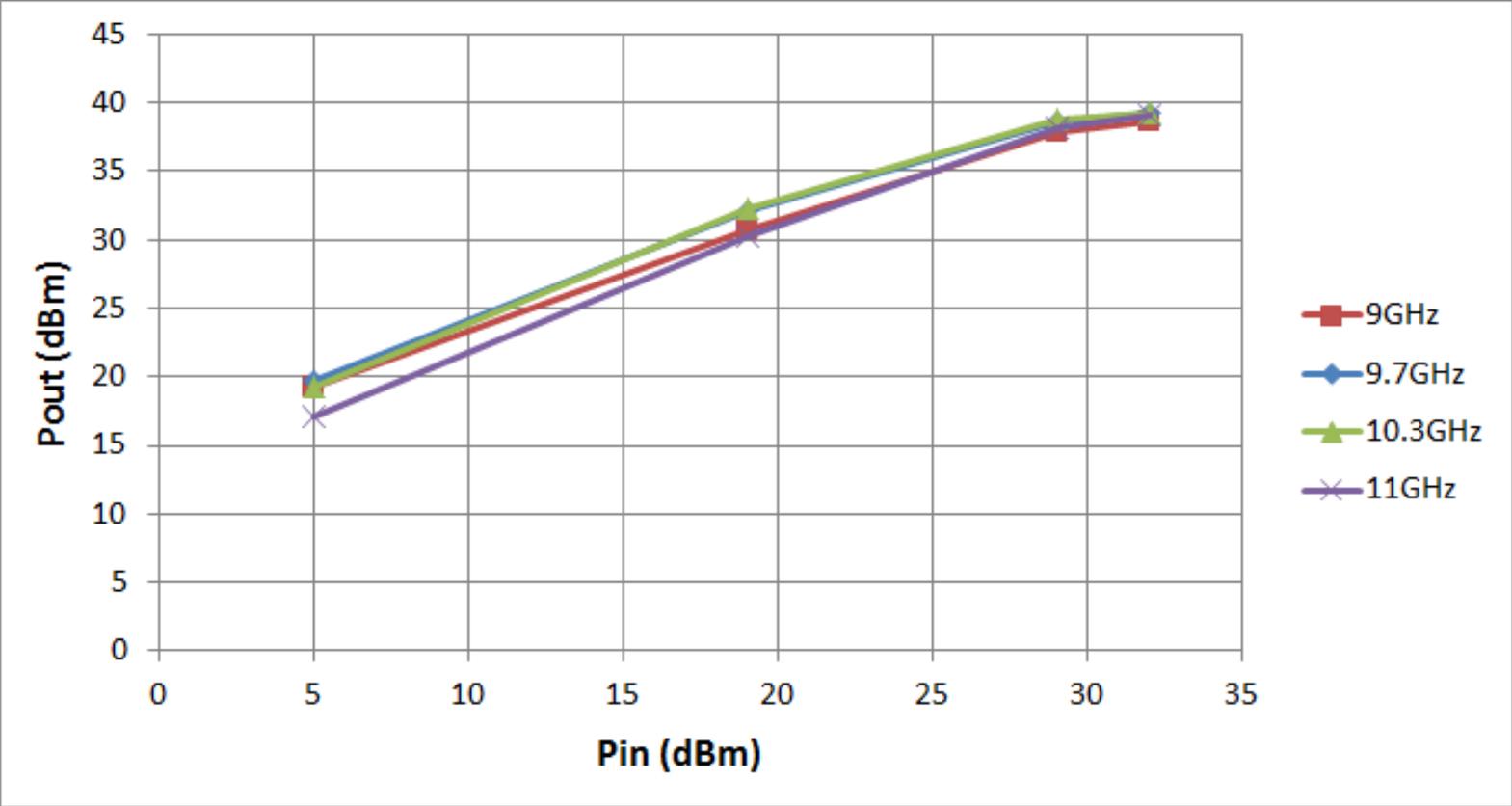


Measured in
blue
Modelled in
red

Quiescent Bias Point = 90mA at 25V

RFOW Measured Power Transfer

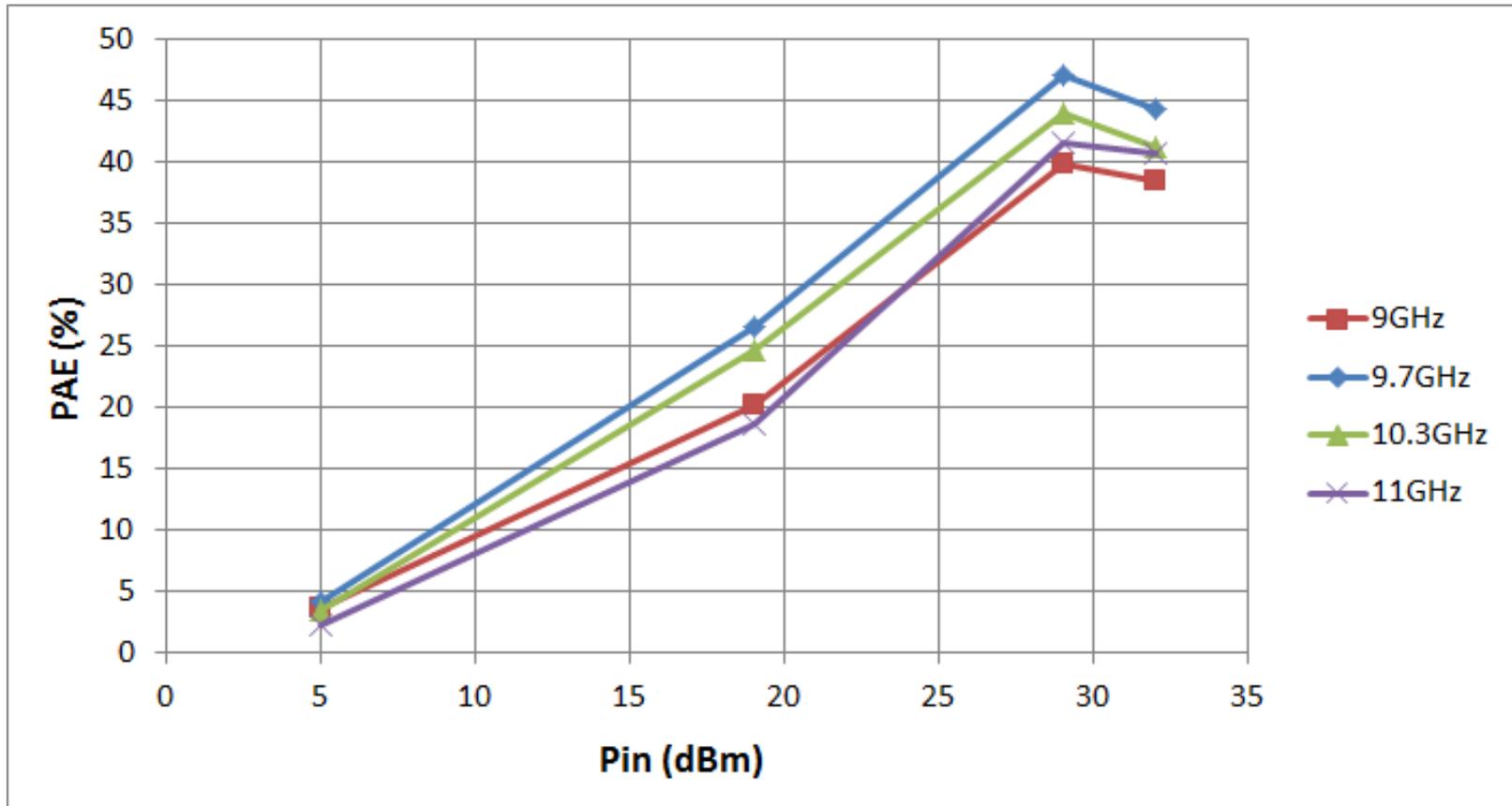
10% Duty Cycle



Quiescent Bias Point = 90mA at 25V

RFOW Measured PAE vs RF Drive

10% Duty Cycle



Quiescent Bias Point = 90mA at 25V

Measured Performance Summary

Quiescent bias, 25V and 90mA

Parameter	Units	Measured	Comment
Frequency	GHz	9 to 11.5	
SS Gain	dB	~ 13	
IRL	dB	> 8	
ORL	dB	> 8	
P_{sat}	dBm	~ 7W (38.5dBm)	+29dBm drive
PAE at Psat	%	~ 42	+29dBm drive
LS gain flatness	dB	± 0.5	

Conclusion

- How to Design an X-Band MMIC PA
 - Thorough device level simulations
 - Careful choice of matching network topology
 - Thorough EM simulations to optimize the layout
- Thankyou for attending
- For more information please see our website
 - www.plextekrfi.com
- Any Questions?

You're Invited

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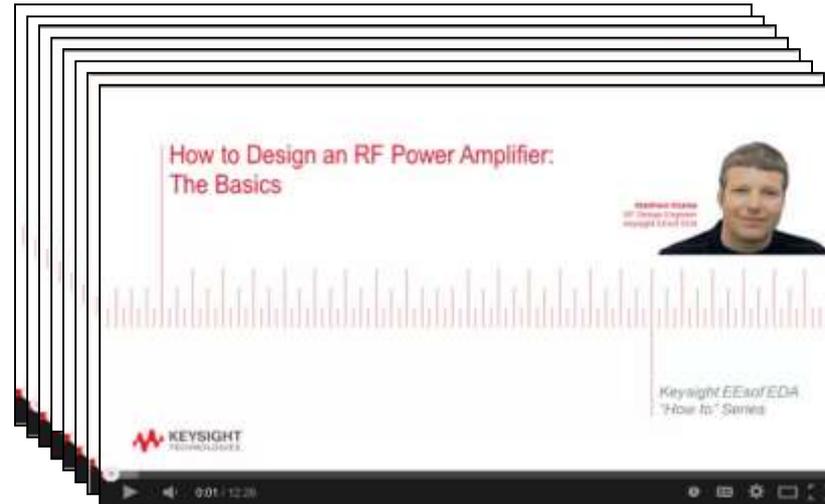
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